



Modeling and control of nonlinear oscillations in neuroscience and energy systems

Anatoly Zlotnik

(with Raphael Nagao, Jr-Shin Li, Istvan Kiss, and Walter Bomela)

Applied Mathematics & Plasma Physics
Los Alamos National Laboratory



Siam Annual Meeting, Portland, OR
July 9, 2018



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2 Phase-Selective Entrainment in Oscillator Assemblies

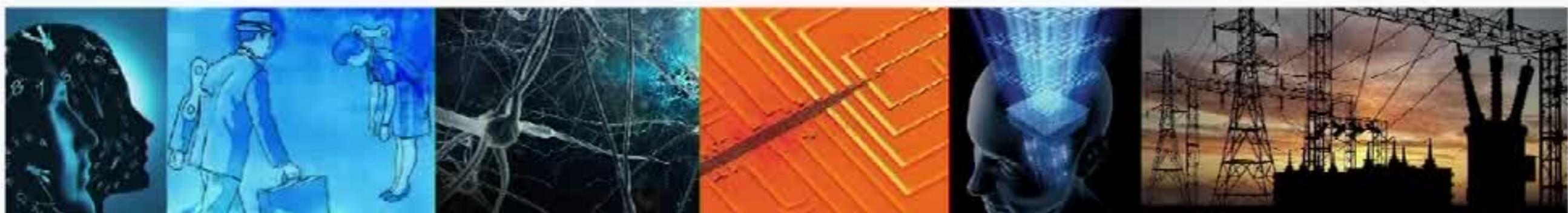
3 Demand Response for Power Distribution

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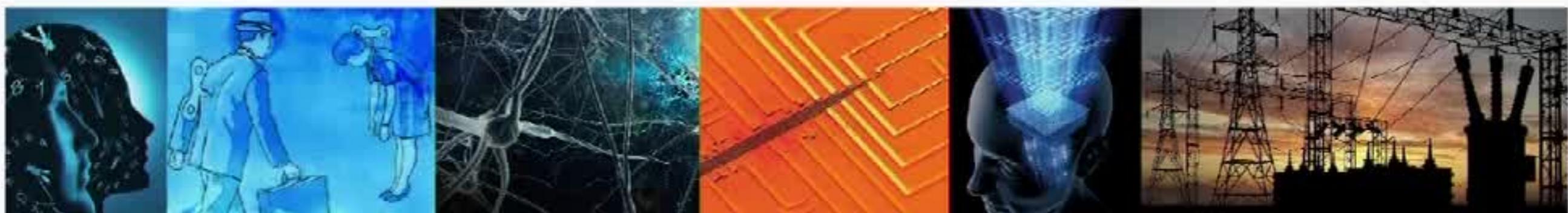
3 Demand Response for Power Distribution

Oscillating Systems in Science and Engineering



- **Natural systems:** spiking of neurons, metabolic chemical reaction systems, circadian timekeepers, cardiac rhythms, auditory processing
- **Medical applications:** deep brain stimulation for Parkinson's disease and epilepsy, cardiac pacemaking, tinnitus treatment, protocols for jet lag recovery
- **Engineering applications:** semiconductor lasers, vibrating systems, power electronics, electrochemical reactors, neurocomputers, power grid design

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Coupled Oscillator Models

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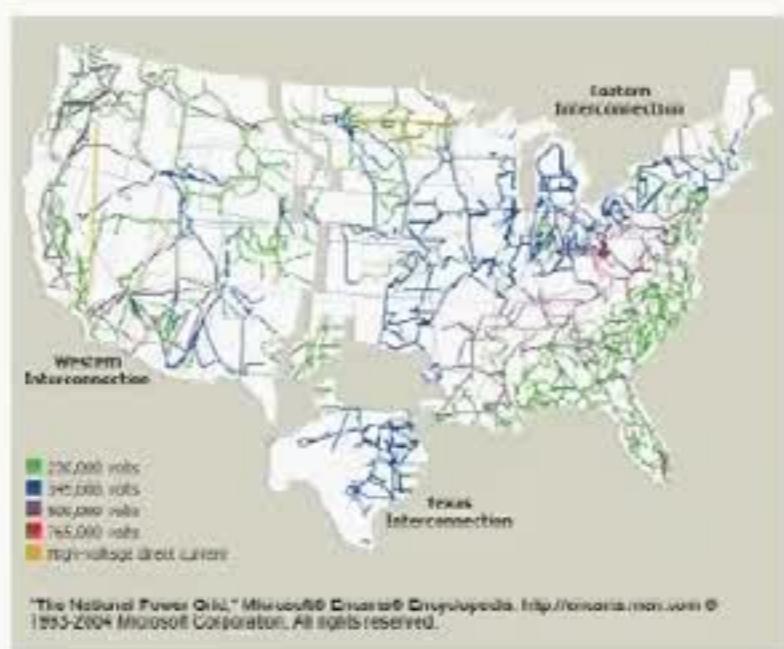
- Firefly synchronization (Buck 1937, Strogatz 1990), Kuramoto model (1975)



$$\dot{\varphi}_i = \omega_i + \frac{K}{N} \sum_{j=1}^N \sin(\varphi_i - \varphi_j)$$

$$r = e^{i\theta} = \frac{1}{N} \sum_{j=1}^N e^{i\varphi_j}$$

- Stability analysis in power networks (Bergen & Hill 1981, Dörfler & Chertkov 2013)



$$D_i \dot{\varphi}_i + P_{1,i} = \sum_{j=1}^N a_{ij} \sin(\varphi_j - \varphi_i), \quad \text{load}$$

$$M_i \ddot{\varphi}_i + D_i \dot{\varphi}_i = P_{m,i} + \sum_{j=1}^N a_{ij} \sin(\varphi_j - \varphi_i), \quad \text{generator}$$

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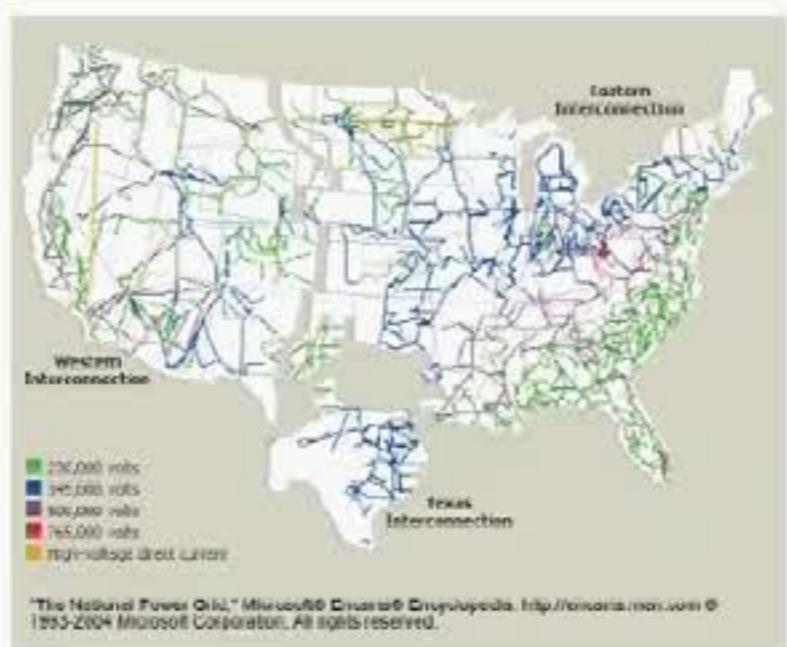
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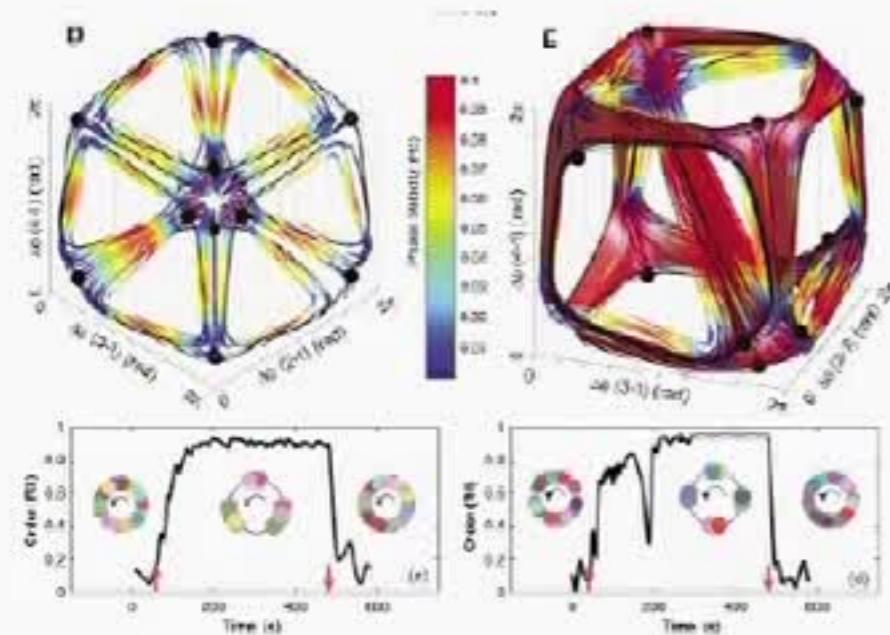


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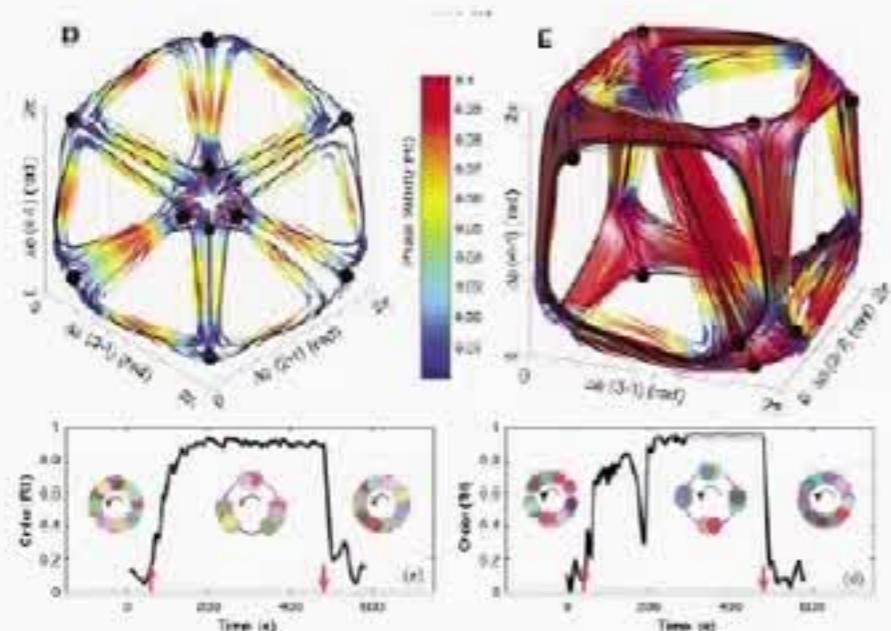
$$H(\varphi_i - \varphi_j) = \frac{1}{2\pi} \int_0^{2\pi} Z(\varphi_i - \varphi_j + \theta) p(\theta) d\theta$$

$$p(t) = \frac{K}{N} \sum_{j=1}^N h(x_j(t)), \quad h(x) = \sum_{n=0}^S k_n x(t - \tau_n)^n$$

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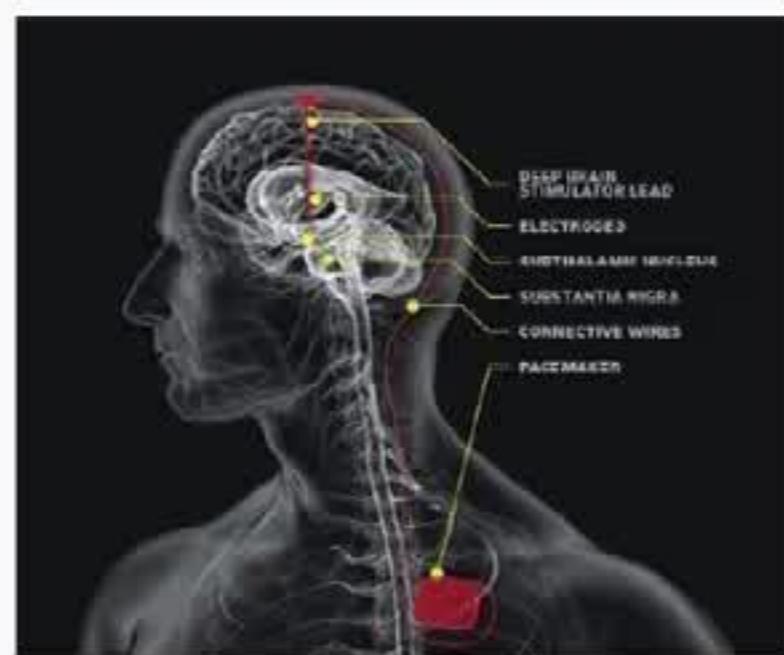


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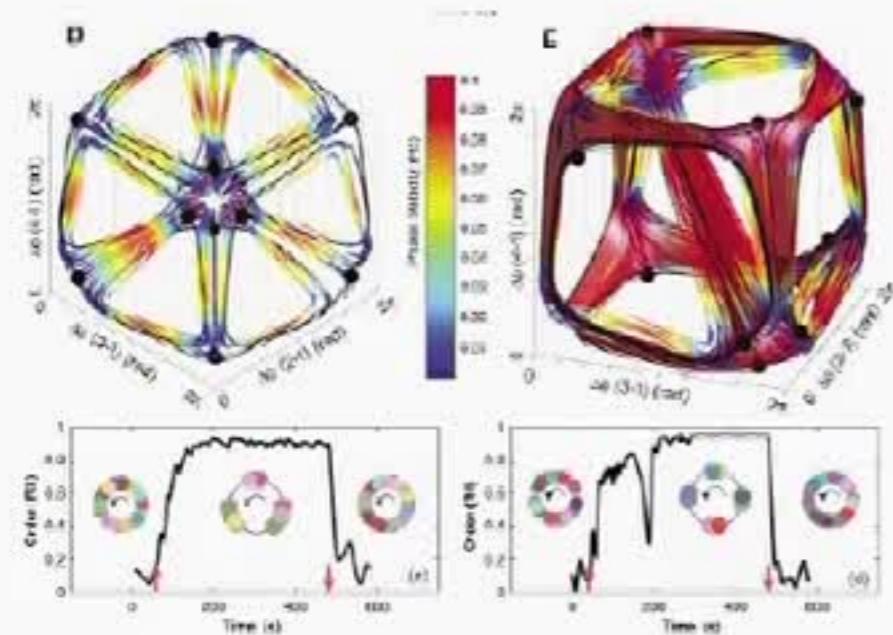


$$\dot{x}_i = f_i(x_i) + \varepsilon \sum_{j=1}^N g_{ij}(x_i) \delta(t - t_j^* - \eta_{ij})$$

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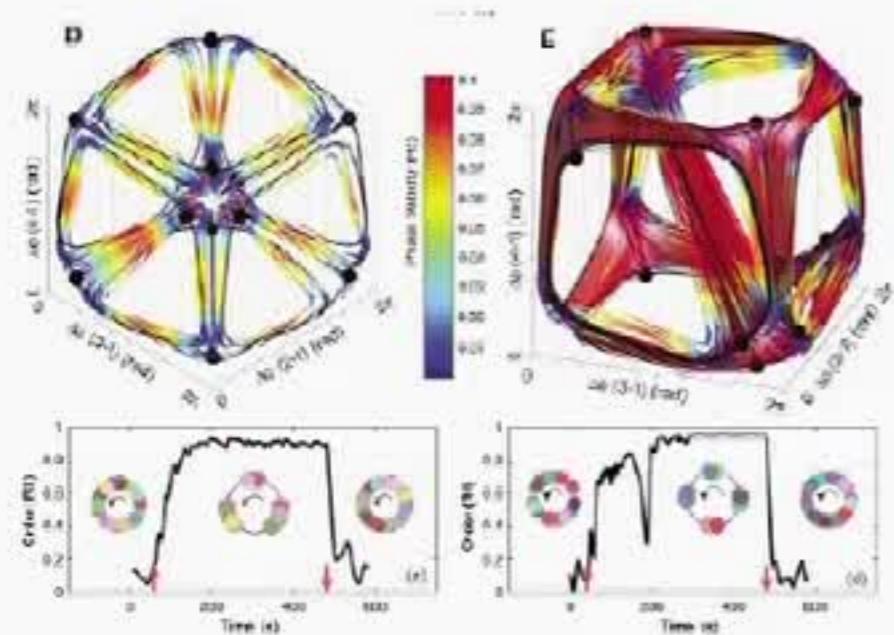
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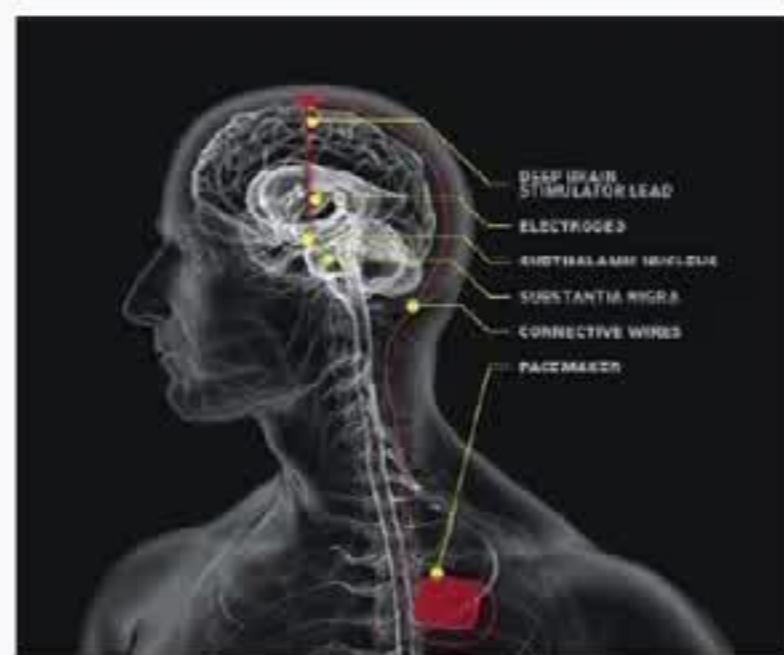


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Synchronization Engineering

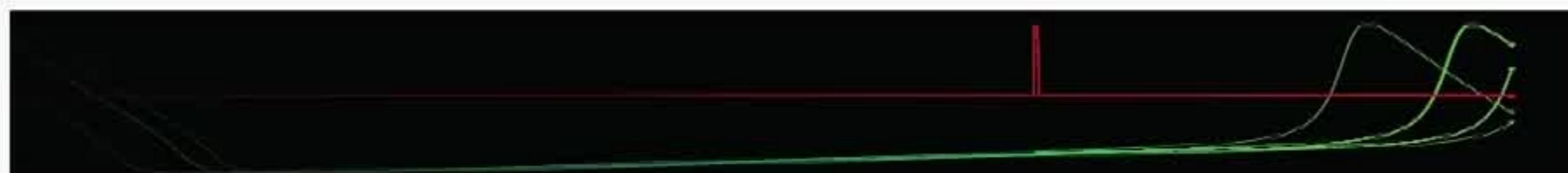
- ***The design of coherent, spatially organized patterns in collections of heterogenous, nonlinear oscillators under the action of weak periodic inputs***
- **A Challenging Problem:**
 - Phase assignment:*** stable pattern in phases of entrained oscillators relative to the forcing phase
 - Robustness:*** stability of phase structure with minimal sensitivity to model parameter variation and unknown disturbances
 - No feedback:*** state information is unavailable or intractable
- **Entrainment:** the dynamic synchronization of one or more oscillators to an external periodic forcing signal

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- **Goal:** robust inputs for nondestructive phase desynchronization of underactuated, noisy and uncertain ensembles of oscillators that cannot be individually observed

Asymptotic Response to a Weak Periodic Input

- Input $u(t) = v(\Omega t)$ where $\theta = \Omega t$ is the **forcing phase** and v is 2π -periodic
- When does $u(t) = v(\Omega t)$ entrain an oscillator $\dot{\psi} = \omega + Z(\psi)u$?
- Entrainment occurs when $\phi(t) = \psi(t) - \Omega t \rightarrow c$
- **Phase drift dynamics:** $\dot{\phi}(t) = \Delta\omega + Z(\phi + \Omega t)u$
- Call $\Delta\omega = \omega - \Omega$ the **frequency detuning**.
- Define an **averaging operator** by

$$\langle v \rangle = \frac{1}{2\pi} \int_0^{2\pi} v(\theta) d\theta$$

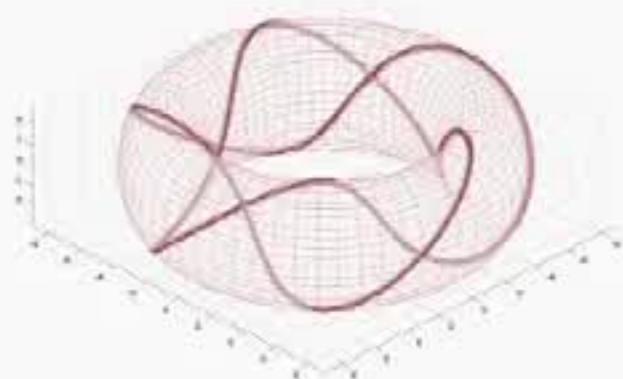
- Make the **weak forcing assumption** $u = \varepsilon u_1$ with $\varepsilon \ll 1$ where $\langle u_1^2 \rangle = 1$.
- The weak ergodic theorem for measure-preserving dynamical systems on the torus and formal averaging eliminate dependence on time
- **Kuramoto:** If forcing is weak, $\phi(t)$ is nearly constant over a single period $[0, T]$

$$\dot{\phi} = \Delta\omega + Z(\phi + \Omega t)f(\Omega t) \approx \Delta\omega + \frac{1}{T} \int_0^T Z(\phi + \Omega t)f(\Omega t) dt = \Delta\omega + \langle Z(\theta + \phi)v(\theta) \rangle$$

- Examine **averaged phase drift dynamics**

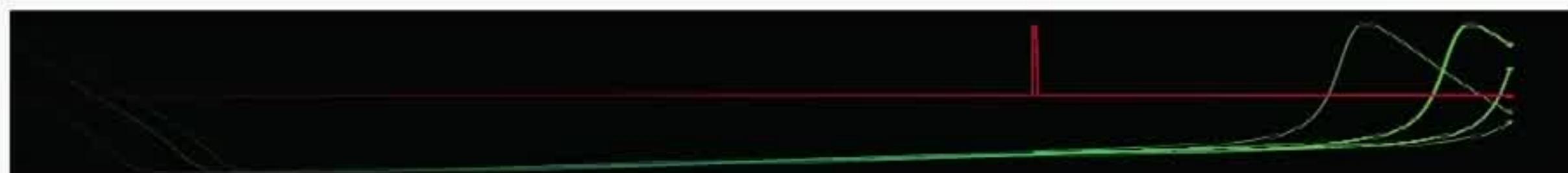
$$\dot{\varphi} = \Delta\omega + \Lambda_v(\varphi)$$

- $\Lambda_v(\varphi) = \langle Z(\theta + \varphi)v(\theta) \rangle$ is the **interaction function**



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Example: Hodgkin-Huxley Spiking Neuron Model

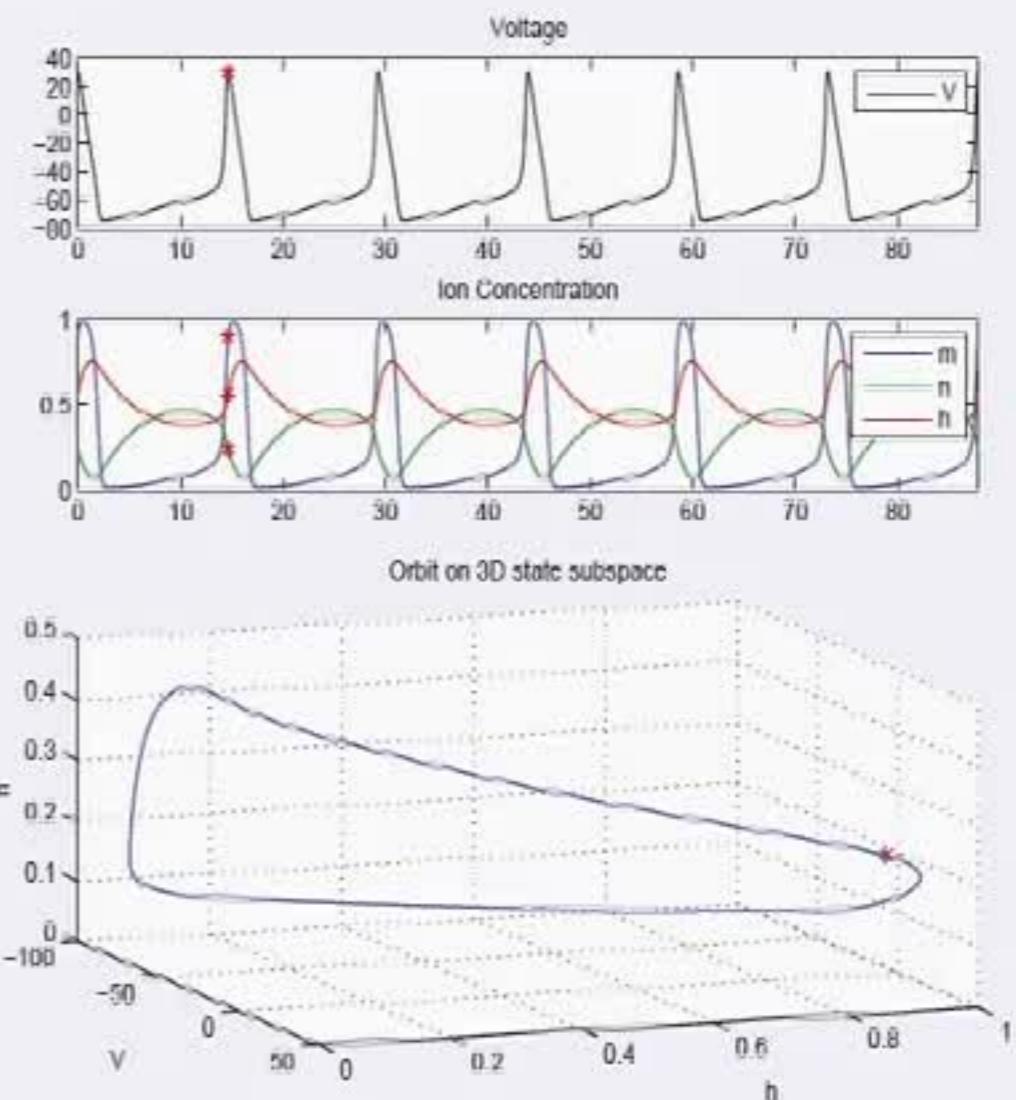
Example: Hodgkin-Huxley Equations (1952)

Model of action potential propagation in a squid giant axon:

$$\begin{aligned} c\dot{V} &= I_b + I(t) - \bar{g}_{Na} h(V - V_{Na})m^3 - \bar{g}_K(V - V_K)n^4 \\ &\quad \dots - \bar{g}_L(V - V_L) \\ \dot{m} &= a_m(V)(1 - m) = b_m(V)m, \\ \dot{h} &= a_h(V)(1 - h) - b_h(V)h, \\ \dot{n} &= a_n(V)(1 - n) - b_n(V)n, \\ a_m(V) &= 0.1(V + 40)/(1 - \exp(-(V + 40)/10)), \\ b_m(V) &= 4 \exp(-(V + 65)/18), \\ a_h(V) &= 0.07 \exp(-(V + 65)/20), \\ b_h(V) &= 1/(1 + \exp(-(V + 35)/10)), \\ a_n(V) &= 0.01(V + 55)/(1 - \exp(-(V + 55)/10)), \\ b_n(V) &= 0.125 \exp(-(V + 65)/80). \end{aligned}$$

$$\begin{aligned} V_{Na} &= 50 \text{ mV}, V_K = -77 \text{ mV}, V_L = -54.4 \text{ mV}, \\ \bar{g}_{Na} &= 120 \text{ mS/cm}^2, \bar{g}_K = 36 \text{ mS/cm}^2, \\ \bar{g}_L &= 0.3 \text{ mS/cm}^2, I_b = 10 \mu\text{A/cm}^2, c = 1 \mu\text{F/cm}^2. \end{aligned}$$

Here $x = (V, m, h, n)$ and $u = I$ are the state and control, respectively.



Minimum energy entrainment of a single oscillator

- Entrain an oscillator $\dot{\psi} = \omega + Z(\psi)u$ to a frequency Ω with minimum energy $\langle v^2 \rangle$
- Requires $\Delta\omega + \Lambda(\varphi_+) \geq 0$ and $\Delta\omega + \Lambda(\varphi_-) \leq 0$, so the active constraint is $\Delta\omega + \Lambda(\varphi_+) = 0$ if $\Omega > \omega$, or $\Delta\omega + \Lambda(\varphi_-) = 0$ if $\Omega < \omega$.
- **Adjoin frequency constraint to minimum energy objective** by Lagrange multiplier λ :

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$$v(\theta) = -\frac{\Delta\omega}{\langle Z^2 \rangle} Z(\theta)$$

- This control is a **re-scaling of the phase response curve**

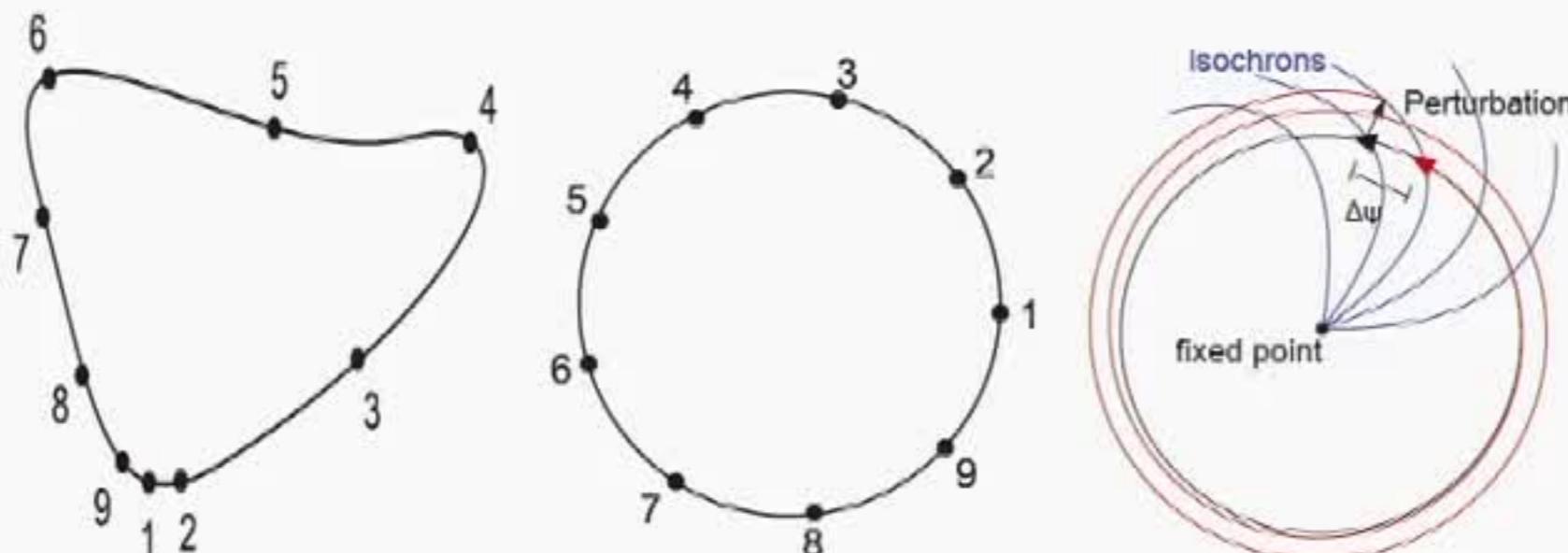
Oscillating Systems and Phase Reduction

- In general, oscillation can refer to any repetitive activity.
- Oscillator:** A smooth dynamical system

$$\dot{x} = f(x, u)$$

where $x \in \mathbb{R}^n$ and $u \in \mathbb{R}$ are the state and control

- An attractive, non-constant limit cycle $\gamma(t) = \gamma(t + T)$ satisfying $\dot{\gamma} = f(\gamma, 0)$ on the periodic orbit $\Gamma = \{y \in \mathbb{R}^n : y = \gamma(t), t \in [0, T]\}$.



State space model	Phase model
$\dot{x} = f(x, u)$	$\dot{\psi} = \omega + Z(\psi)u$
State: $x \in \mathbb{R}^n$	phase: $\psi \in [0, 2\pi)$
control: $u \in \mathbb{R}$	

- $\psi \equiv$ asymptotic phase
- $\omega \equiv$ natural frequency
- $Z \equiv$ **phase response curve (PRC)**: infinitesimal sensitivity to input u at a given phase.

Asymptotic Response to a Weak Periodic Input

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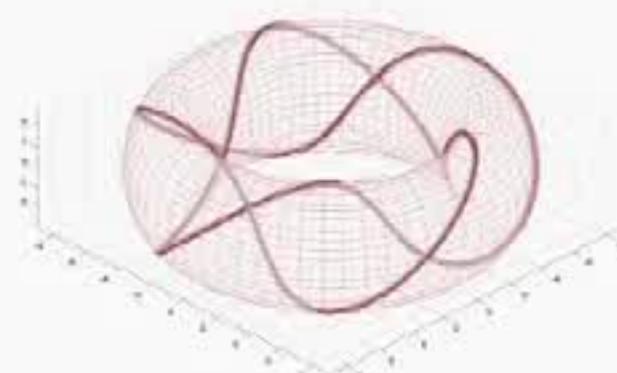
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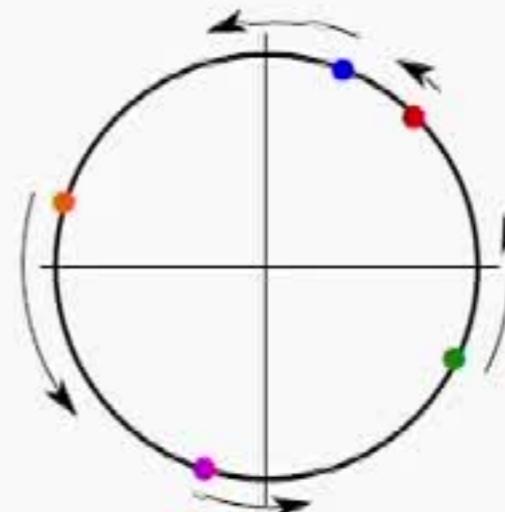
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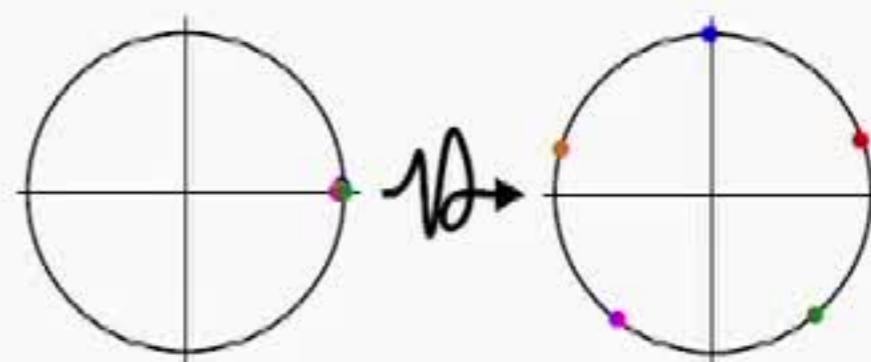


Phase Patterns

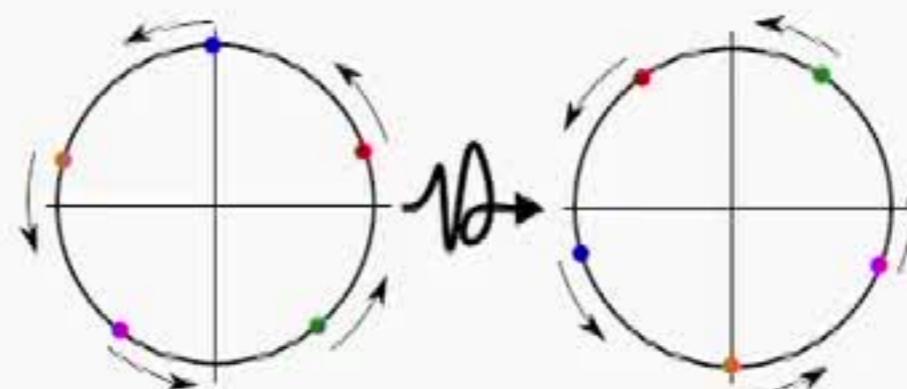
- Collection of phase oscillators
 $\{\dot{\psi}_i = \omega_i + Z(\psi)u\}$ for $i = 1, \dots, P$, with
 $\omega_i \neq \omega_j$ for $i \neq j$.



- **Objective 1:** Establish a fixed phase relationship. If $\psi_i(0)$ is unknown for all $i = 1, \dots, N$, steer the oscillators to $\psi_i(T) = \phi_i^\infty$ where ϕ_i^∞ are desired phases for $i = 1, \dots, P$



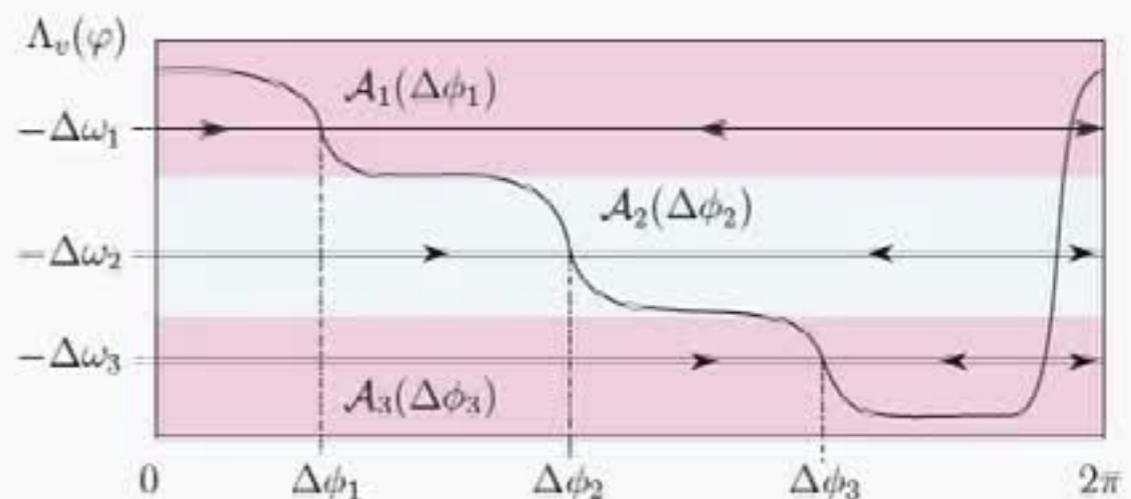
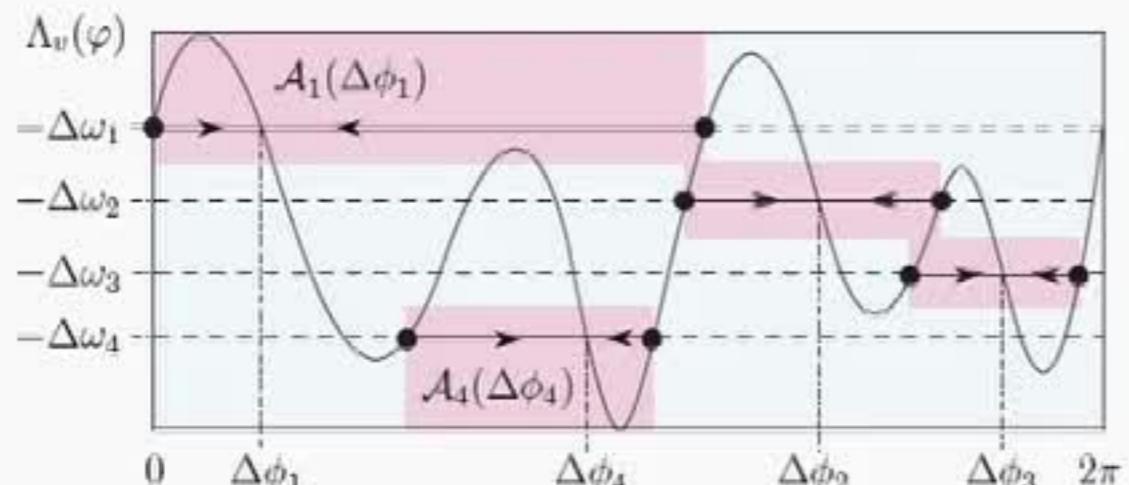
- **Objective 2:** Maintain a fixed phase relationship. After Objective 1 is established, produce a repeated input that ensures $\psi_i(kT) = \phi_i^\infty$ for $i = 1, \dots, N$ for all $k \in \mathbb{N}$.



Interaction Function Design

- Entrain $\mathcal{F} = \{\dot{\psi}_i = \omega_i + Z(\psi_i)u, i = 1, \dots, P\}$ using $u(t) = v(\Omega t)$
- $\phi_i(t) = \psi_i(t) - \Omega t$ are relative to forcing phase $\theta = \Omega t$
- **Goal:** Assign $\phi_i(t) = \Delta\phi_i$
- **Averaging:** $\langle \mathcal{F} \rangle = \{\dot{\varphi}_i = \Delta\omega_i + \Lambda_v(\varphi_i), i = 1, \dots, P\}, \varphi_i^\infty = \lim_{t \rightarrow \infty} \varphi_i(t)$
- **Attractivity:** $\mathcal{A}_i(\varphi_i^\infty) \subset [0, 2\pi]$ is the region that attracts to φ_i^∞
- **Design Criteria:** For $i = 1, \dots, P$,

$$(i) \Lambda_v(\Delta\phi_i) = -\Delta\omega_i, \quad (ii) \Lambda'_v(\Delta\phi_i) < 0, \quad (iii) \mathcal{A}_i(\Delta\phi_i) = [0, 2\pi]$$



- **Key Idea:** Design interaction function, reverse-engineer control using PRC

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Coherent Phase Pattern Control Waveform

- Represent Z and v using truncated Fourier series expansions

$$Z(\theta) \approx Z^r(\theta) = \frac{a_0}{2} + \sum_{n=1}^r [a_n \cos(n\theta) + b_n \sin(n\theta)],$$

$$v(\theta) \approx v^r(\theta) = \frac{c_0}{2} + \sum_{n=1}^r [c_n \cos(n\theta) + d_n \sin(n\theta)],$$

- Series for $\Lambda_v(\varphi) = \langle Z(\theta + \varphi)v(\theta) \rangle$ obtained using trigonometric angle sum identities and the orthogonality of the Fourier basis

$$\Lambda_v^r(\varphi) = \frac{f_0}{2} + \frac{1}{2} \sum_{n=1}^r f_n \cos(n\varphi) + \frac{1}{2} \sum_{n=1}^r g_n \sin(n\varphi),$$

where $f_0 = \frac{a_0 c_0}{2}$, $f_n = a_n c_n + b_n d_n$, $g_n = b_n c_n - a_n d_n$.

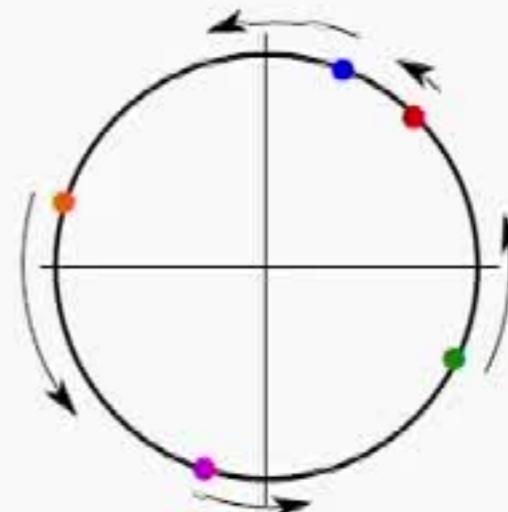
- Coefficients of truncated Fourier series for the control waveform $v^r(\theta)$ are

$$c_0 = 4 \frac{f_0}{a_0} \chi_{[a_0 \neq 0]}, \quad c_n = 2 \frac{(f_n a_n + b_n g_n) \chi_{[a_n^2 + b_n^2 \neq 0]}}{a_n^2 + b_n^2}, \quad d_n = 2 \frac{(f_n b_n - a_n g_n) \chi_{[a_n^2 + b_n^2 \neq 0]}}{a_n^2 + b_n^2},$$

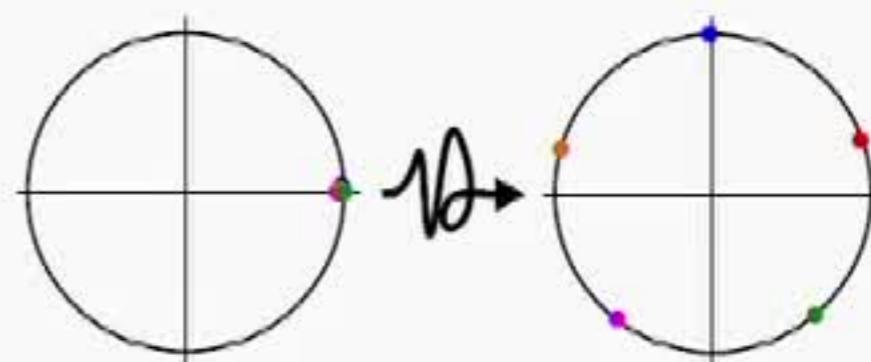
where $\chi_A = 1$ if A is true, and $\chi_A = 0$ otherwise.

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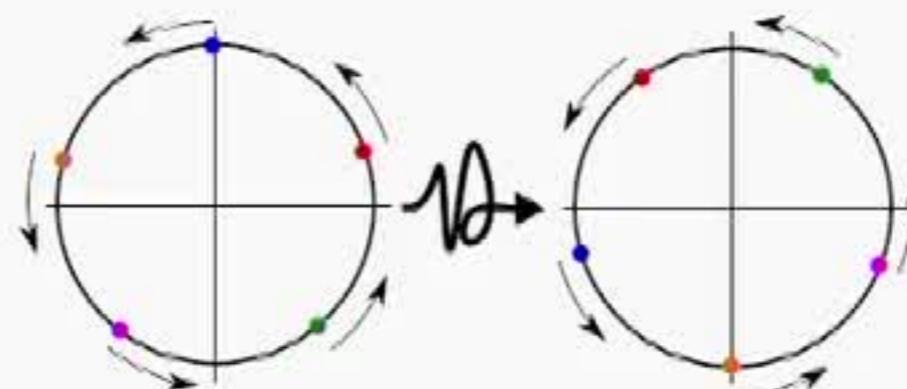
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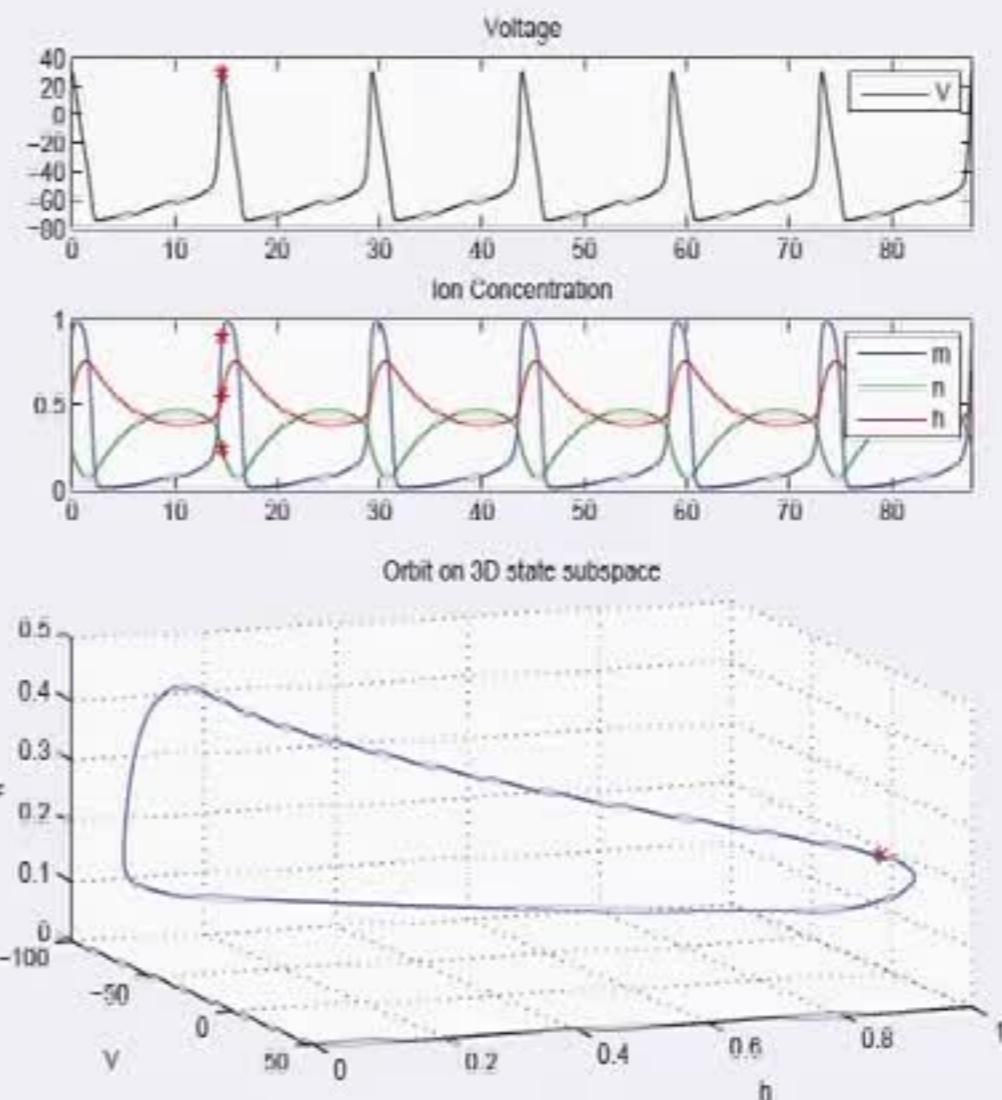
$$\begin{aligned} c\dot{V} &= I_b + I(t) - \bar{g}_{Na} h(V - V_{Na})m^3 - \bar{g}_K (V - V_K)n^4 \\ &\quad \dots - \bar{g}_L (V - V_L) \\ \dot{m} &= a_m(V)(1-m) = b_m(V)m, \\ \dot{h} &= a_h(V)(1-h) - b_h(V)h, \\ \dot{n} &= a_n(V)(1-n) - b_n(V)n, \\ a_m(V) &= 0.1(V + 40)/(1 - \exp(-(V + 40)/10)), \\ b_m(V) &= 4 \exp(-(V + 65)/18), \\ a_h(V) &= 0.07 \exp(-(V + 65)/20), \\ b_h(V) &= 1/(1 + \exp(-(V + 35)/10)), \\ a_n(V) &= 0.01(V + 55)/(1 - \exp(-(V + 55)/10)), \\ b_n(V) &= 0.125 \exp(-(V + 65)/80). \end{aligned}$$

$$V_{Na} = 50 \text{ mV}, V_K = -77 \text{ mV}, V_L = -54.4 \text{ mV},$$

$$\bar{g}_{Na} = 120 \text{ mS/cm}^2, \bar{g}_K = 36 \text{ mS/cm}^2,$$

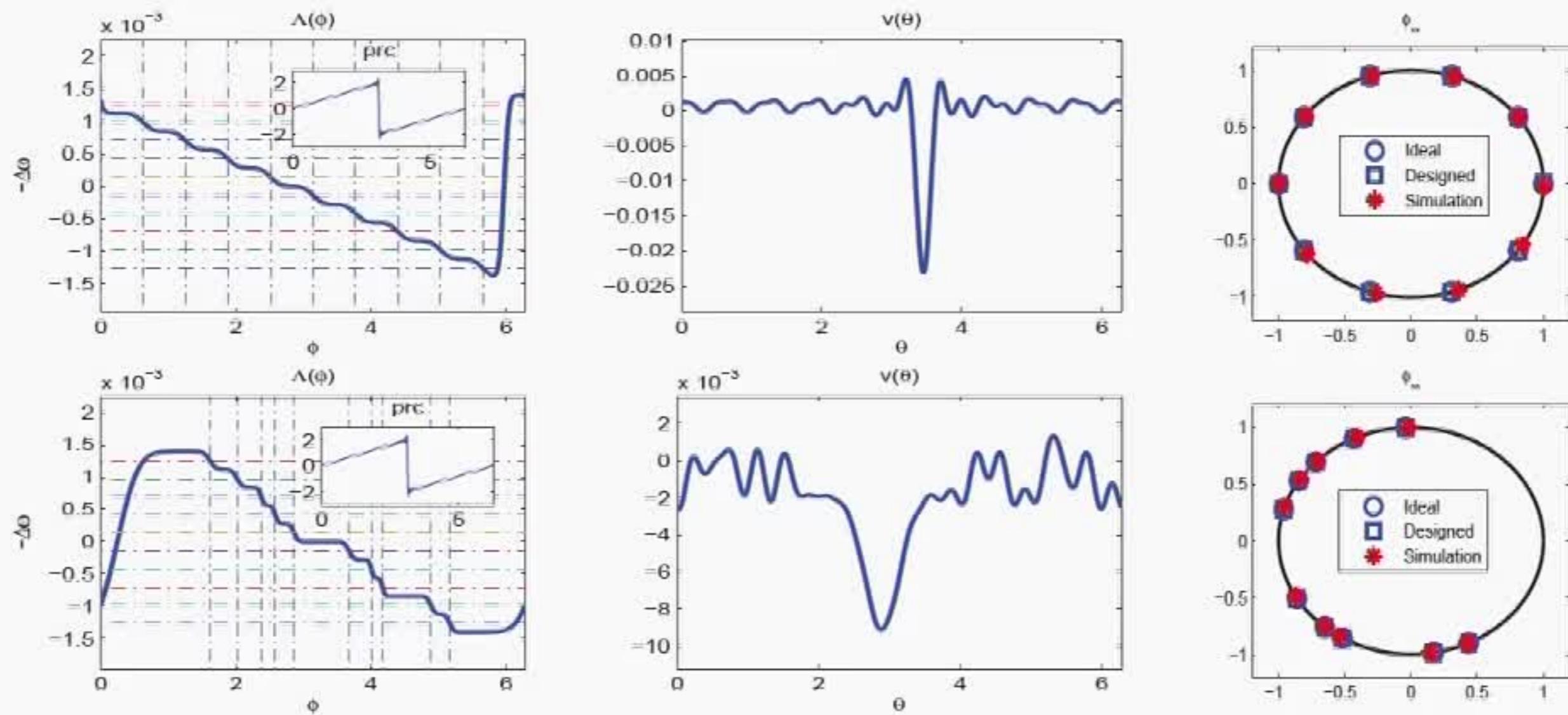
$$\bar{g}_L = 0.3 \text{ mS/cm}^2, I_b = 10 \mu\text{A/cm}^2, c = 1 \mu\text{F/cm}^2.$$

Here $x = (V, m, h, n)$ and $u = I$ are the state and control, respectively.



Pattern Complexity

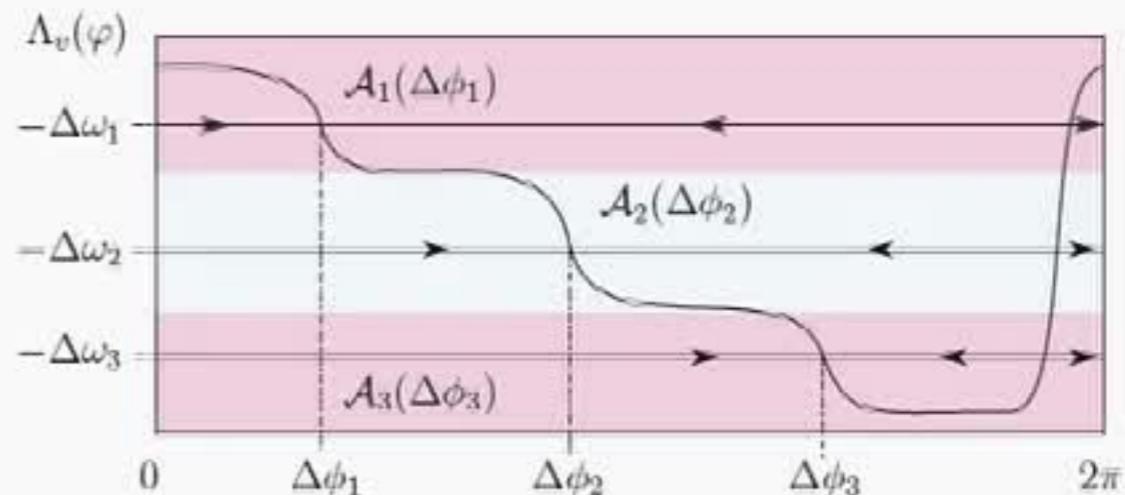
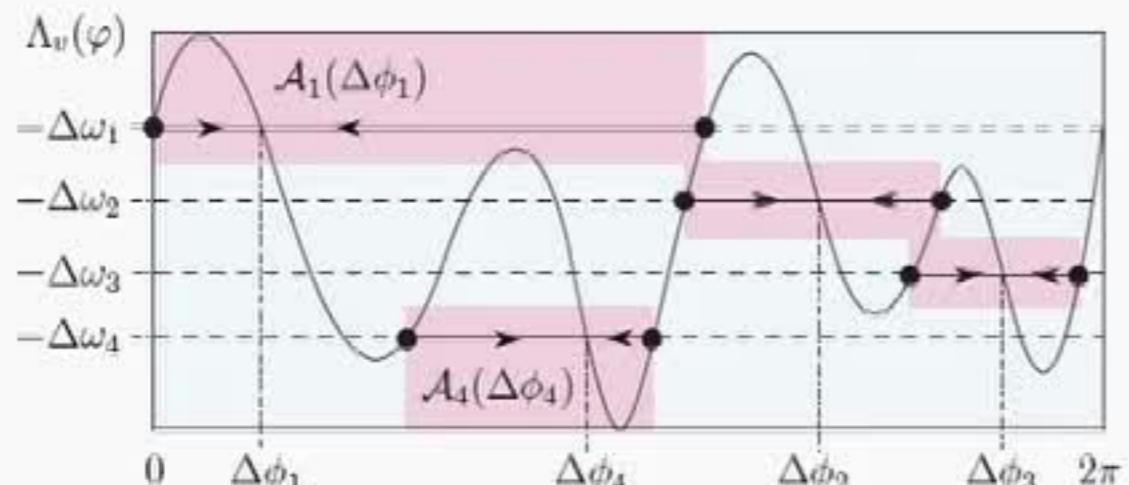
- Complexity of phase patterns is limited by information/complexity in the PRC
- Consider theoretical example of a PRC with sawtooth shape
- Interaction function $\Lambda_v^r(\varphi)$ designed using a $r = 20$ order Fourier series fit
- 10 oscillators with frequencies on $[0.995\omega, 1.005\omega]$, where $\omega = 2\pi/T$ with $T = 25$.



Interaction Function Design

- Entrain $\mathcal{F} = \{\dot{\psi}_i = \omega_i + Z(\psi_i)u, i = 1, \dots, P\}$ using $u(t) = v(\Omega t)$
- $\phi_i(t) = \psi_i(t) - \Omega t$ are relative to forcing phase $\theta = \Omega t$
- **Goal:** Assign $\phi_i(t) = \Delta\phi_i$
- **Averaging:** $\langle \mathcal{F} \rangle = \{\dot{\varphi}_i = \Delta\omega_i + \Lambda_v(\varphi_i), i = 1, \dots, P\}, \varphi_i^\infty = \lim_{t \rightarrow \infty} \varphi_i(t)$
- **Attractivity:** $\mathcal{A}_i(\varphi_i^\infty) \subset [0, 2\pi]$ is the region that attracts to φ_i^∞
- **Design Criteria:** For $i = 1, \dots, P$,

$$(i) \Lambda_v(\Delta\phi_i) = -\Delta\omega_i, \quad (ii) \Lambda'_v(\Delta\phi_i) < 0, \quad (iii) \mathcal{A}_i(\Delta\phi_i) = [0, 2\pi]$$



- **Key Idea:** Design interaction function, reverse-engineer control using PRC

Coherent Phase Pattern Control Waveform

- Represent Z and v using truncated Fourier series expansions

$$Z(\theta) \approx Z^r(\theta) = \frac{a_0}{2} + \sum_{n=1}^r [a_n \cos(n\theta) + b_n \sin(n\theta)],$$

$$v(\theta) \approx v^r(\theta) = \frac{c_0}{2} + \sum_{n=1}^r [c_n \cos(n\theta) + d_n \sin(n\theta)],$$

- Series for $\Lambda_v(\varphi) = \langle Z(\theta + \varphi)v(\theta) \rangle$ obtained using trigonometric angle sum identities and the orthogonality of the Fourier basis

$$\Lambda_v^r(\varphi) = \frac{f_0}{2} + \frac{1}{2} \sum_{n=1}^r f_n \cos(n\varphi) + \frac{1}{2} \sum_{n=1}^r g_n \sin(n\varphi),$$

where $f_0 = \frac{a_0 c_0}{2}$, $f_n = a_n c_n + b_n d_n$, $g_n = b_n c_n - a_n d_n$.

- Coefficients of truncated Fourier series for the control waveform $v^r(\theta)$ are

$$c_0 = 4 \frac{f_0}{a_0} \chi_{[a_0 \neq 0]}, \quad c_n = 2 \frac{(f_n a_n + b_n g_n) \chi_{[a_n^2 + b_n^2 \neq 0]}}{a_n^2 + b_n^2}, \quad d_n = 2 \frac{(f_n b_n - a_n g_n) \chi_{[a_n^2 + b_n^2 \neq 0]}}{a_n^2 + b_n^2},$$

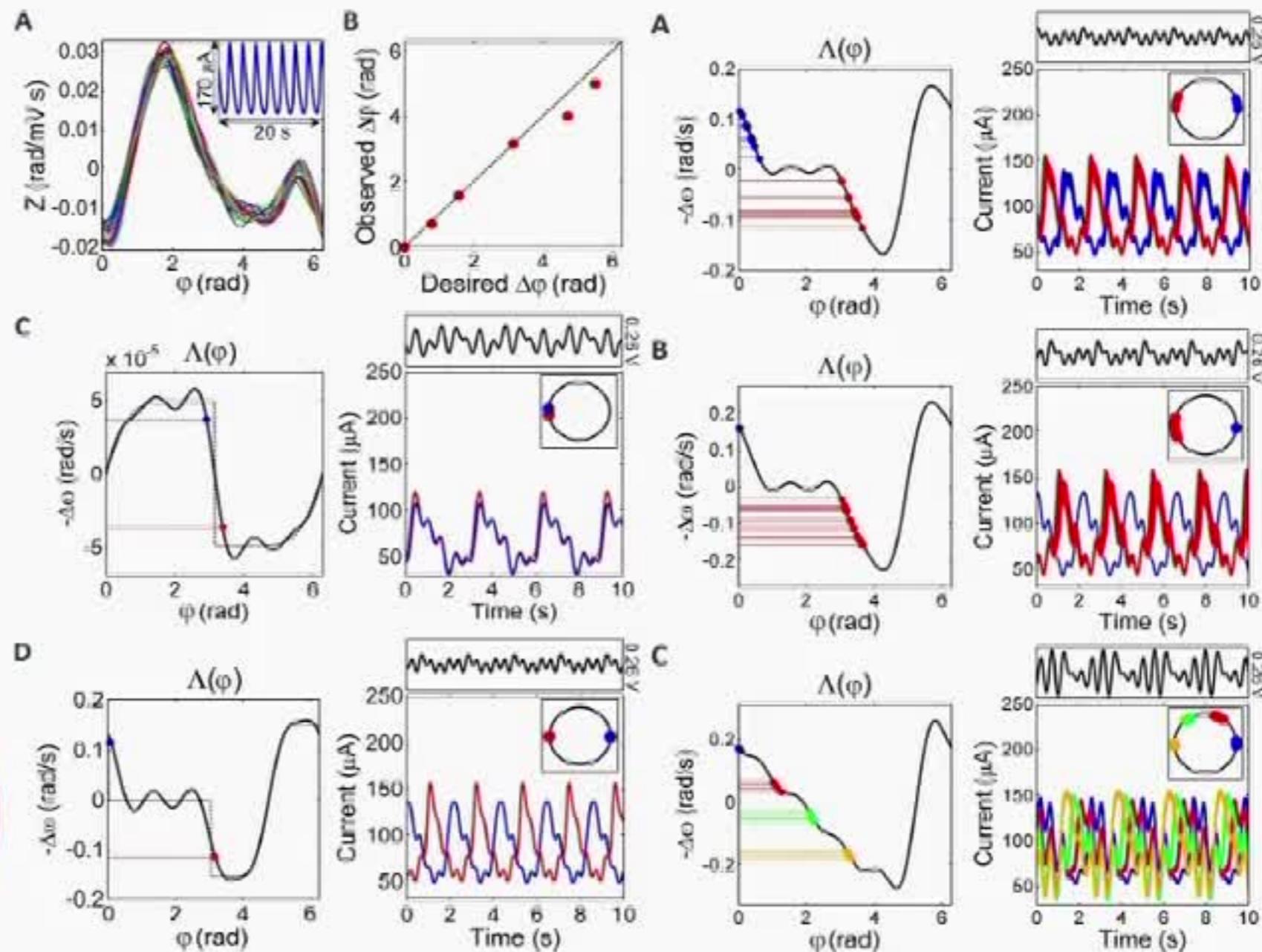
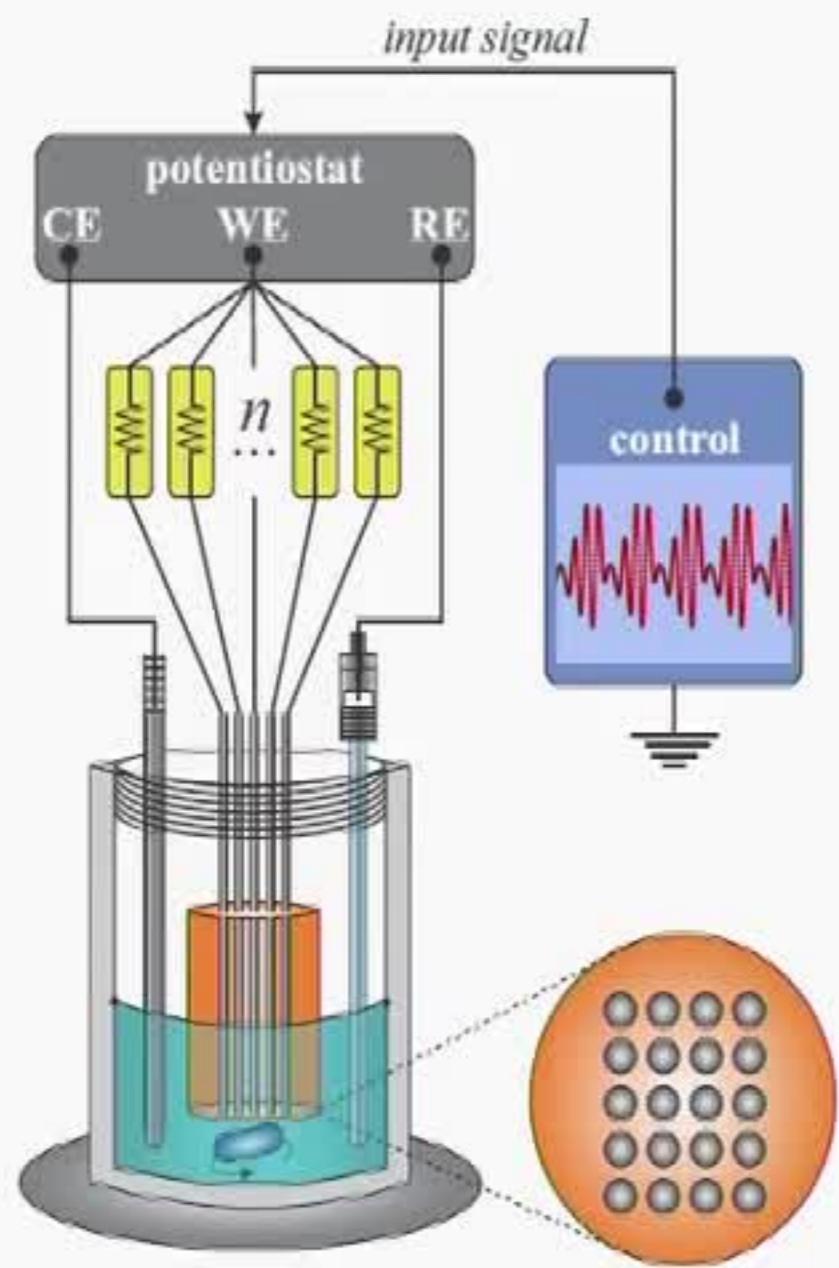
where $\chi_A = 1$ if A is true, and $\chi_A = 0$ otherwise.

Phase Selection with Electrochemical Reactions

- Electrochemical oscillation: dissolution of nickel electrodes in concentrated sulfuric acid solution

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Hodgkin-Huxley Spiking Neuron Model

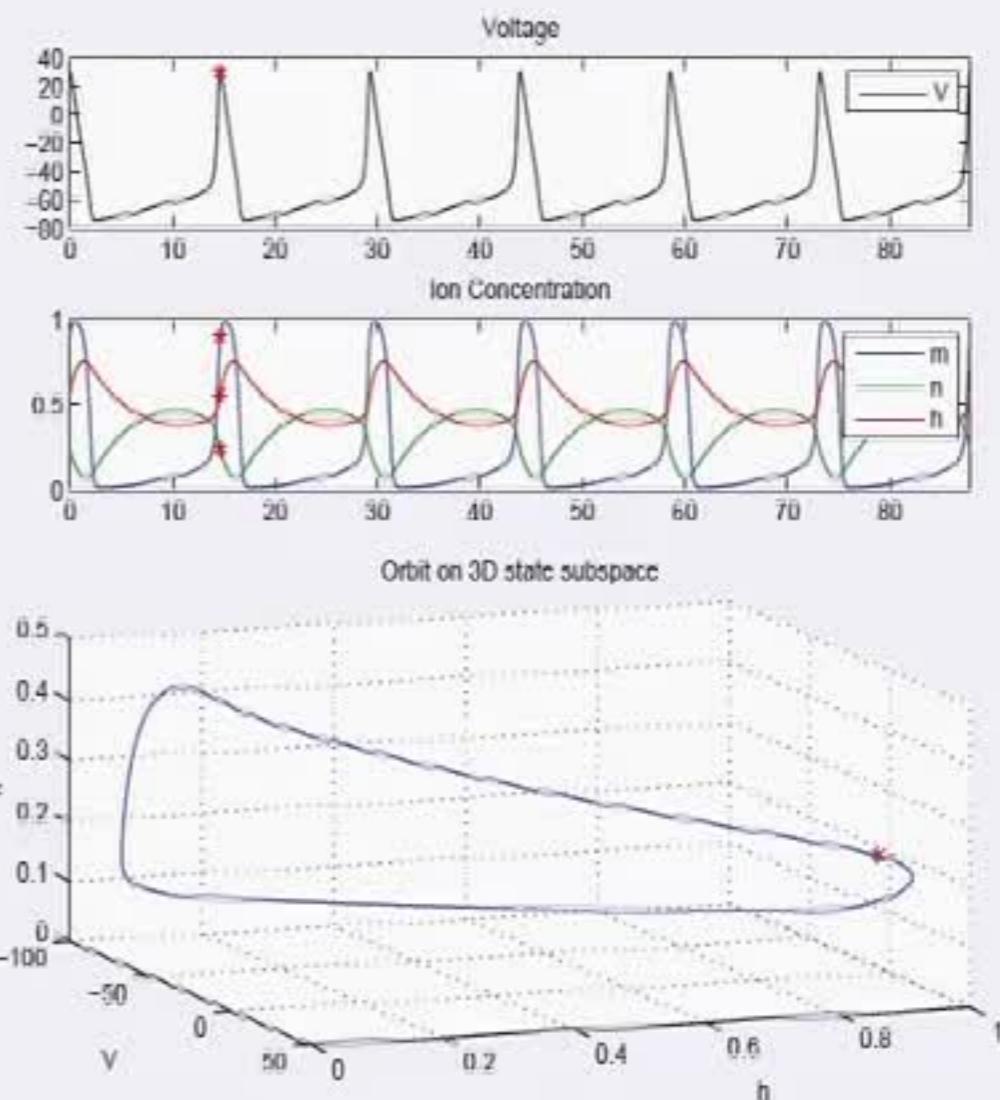
Example: Hodgkin-Huxley Equations (1952)

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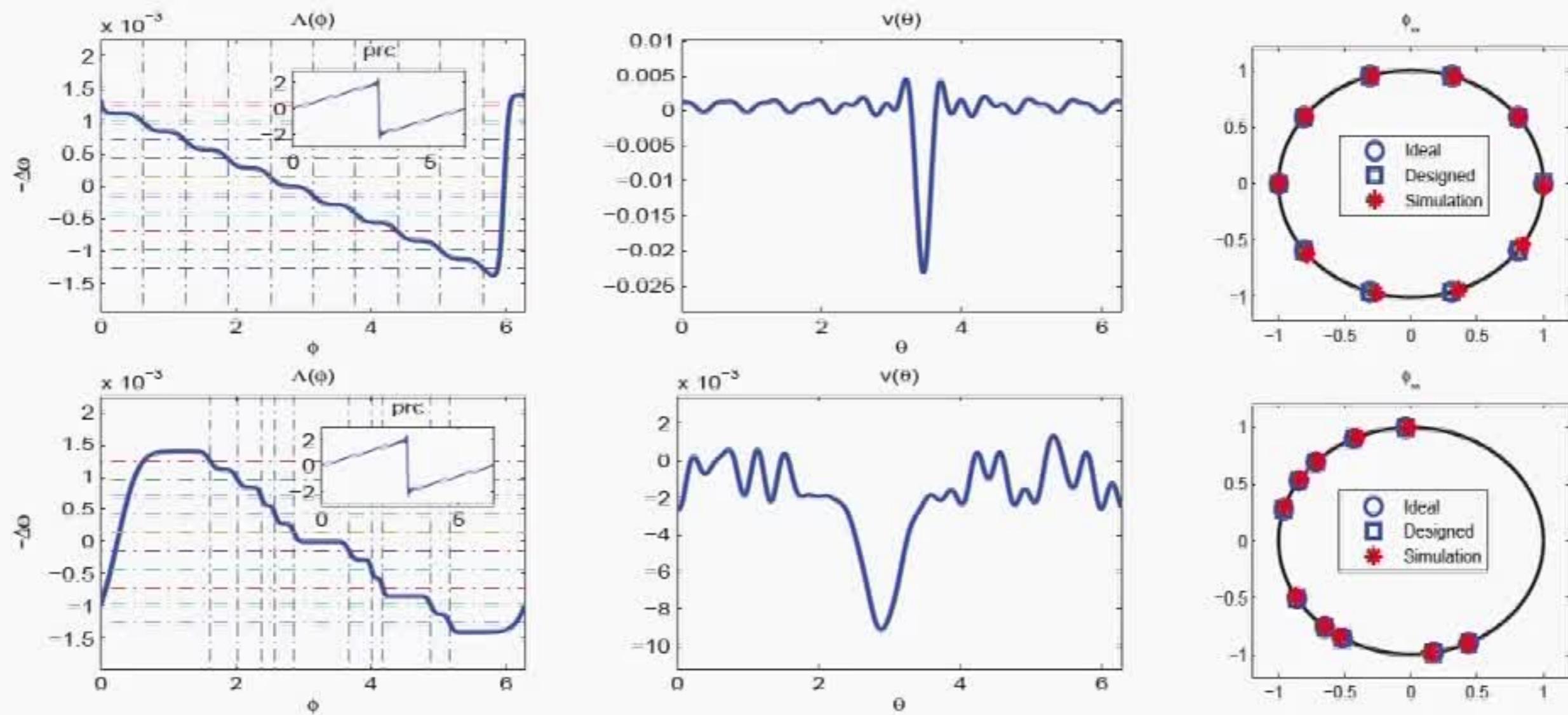
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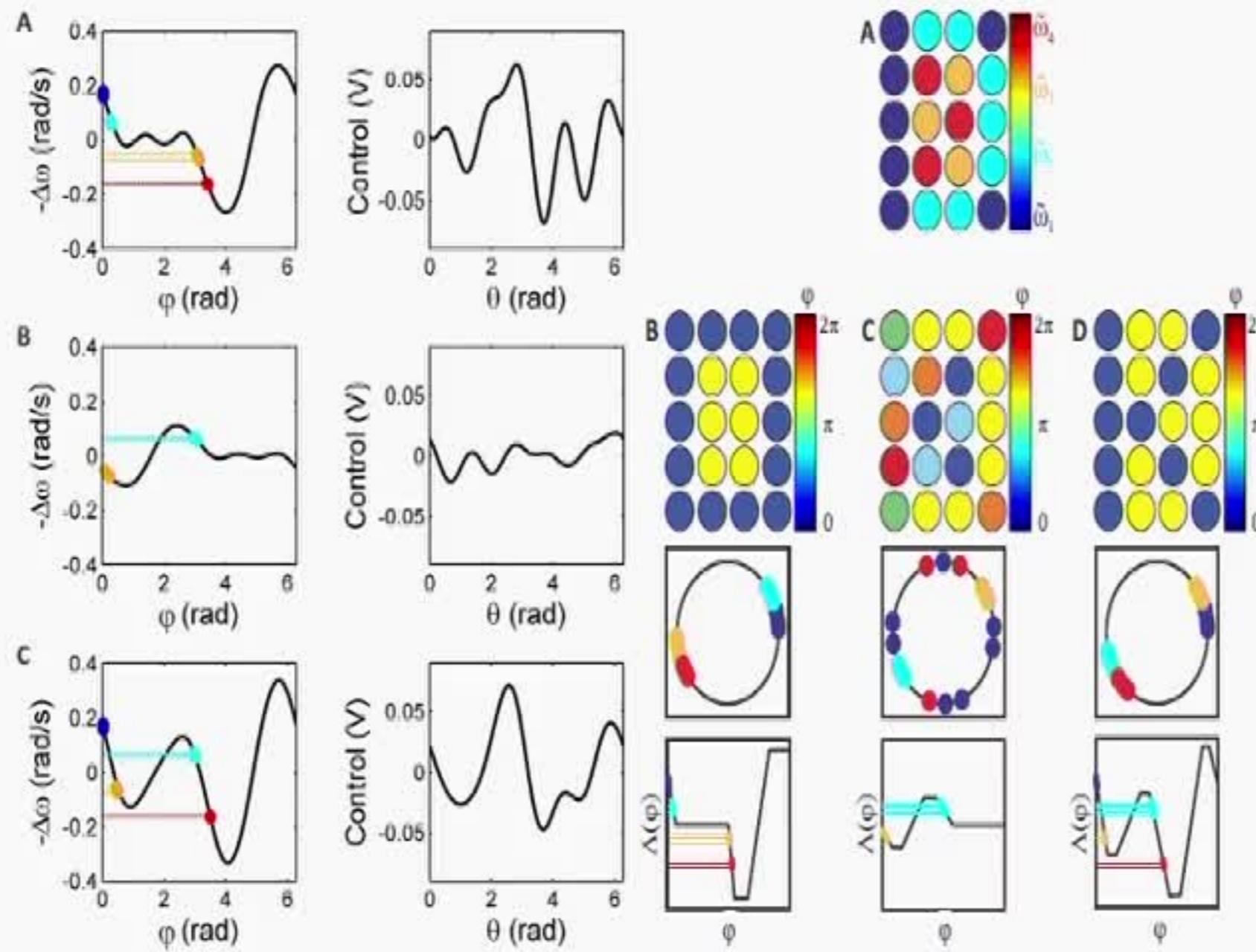
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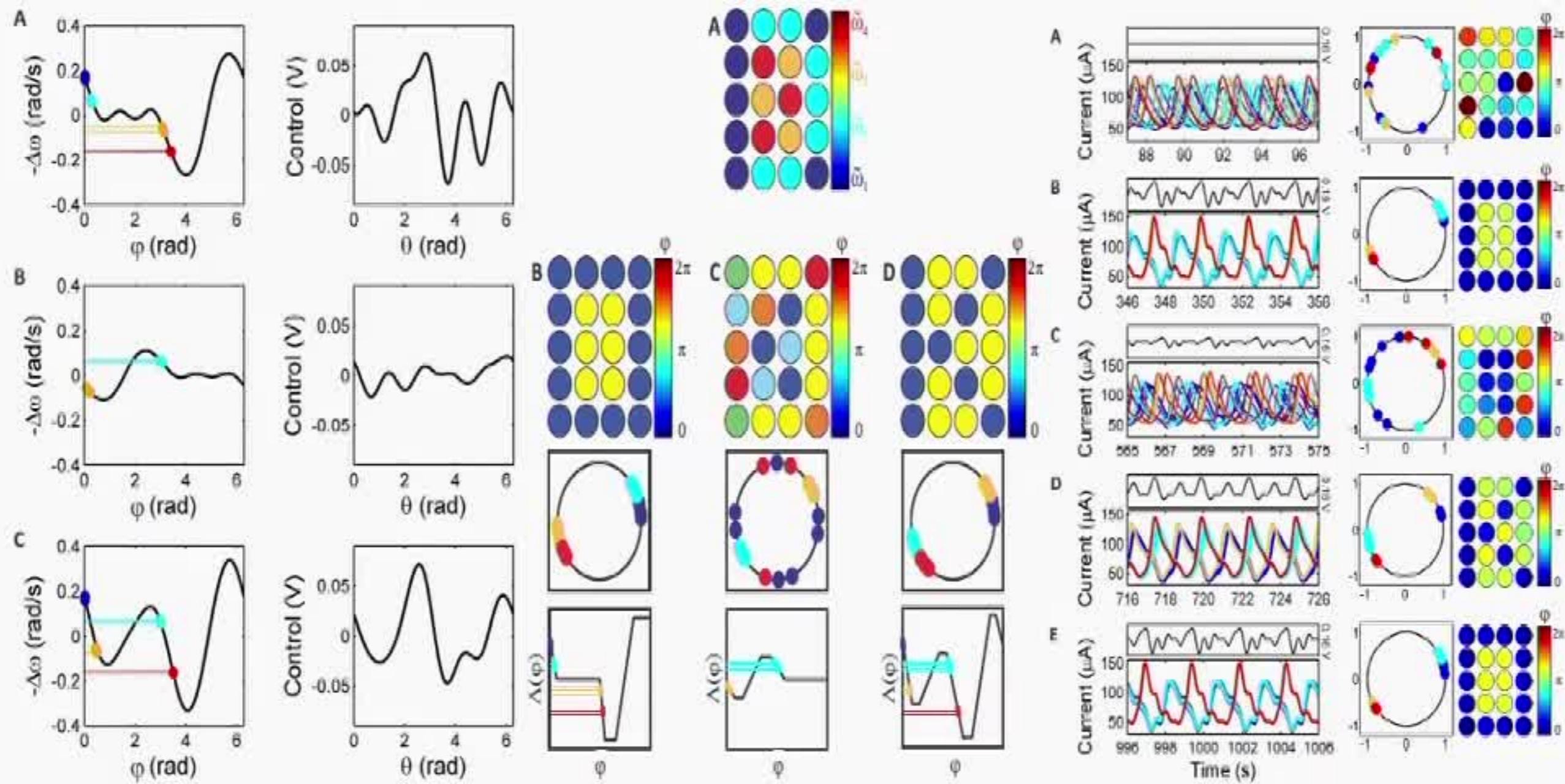
Phase Selection with Electrochemical Reactions

■ Pattern switching control



Phase Selection with Electrochemical Reactions

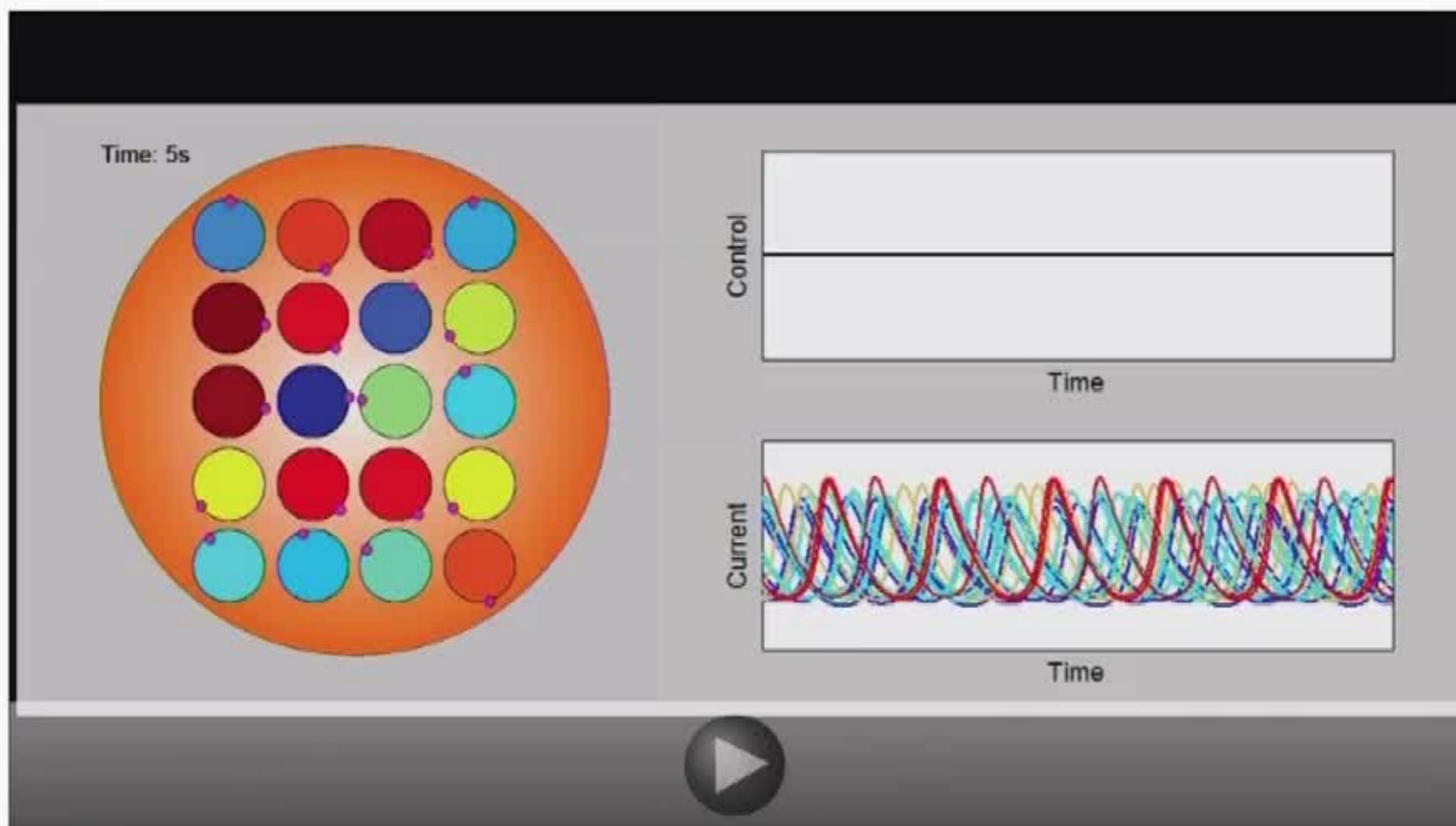
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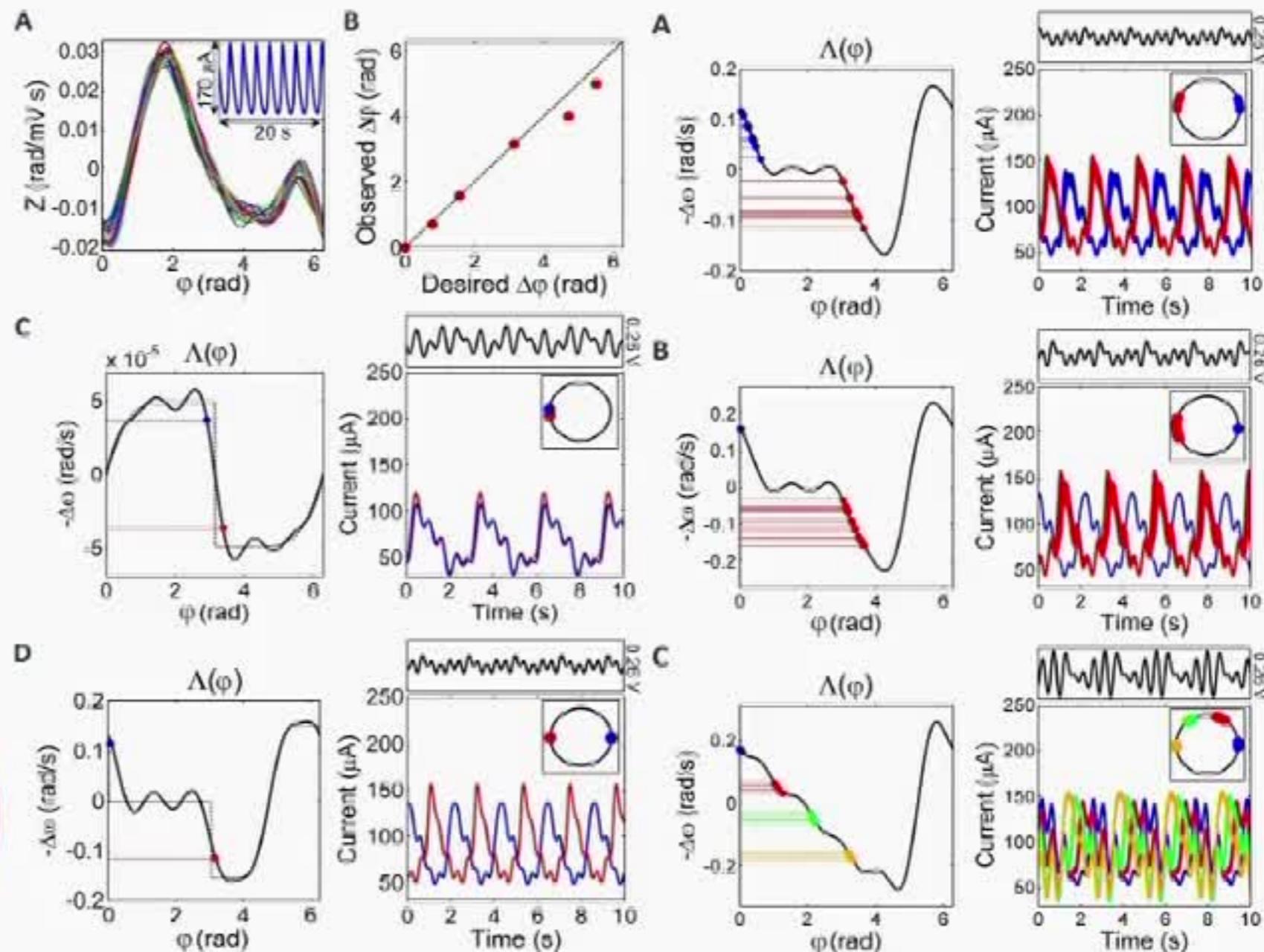
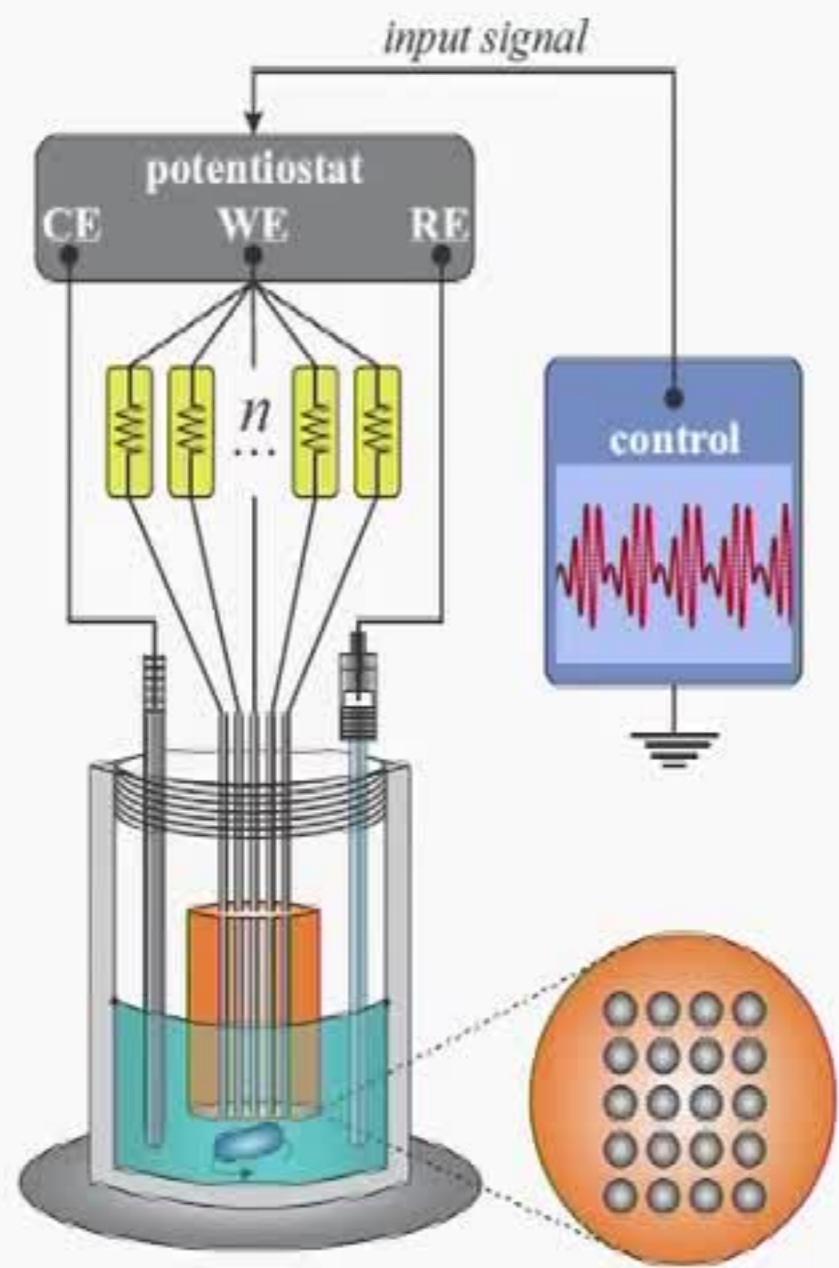
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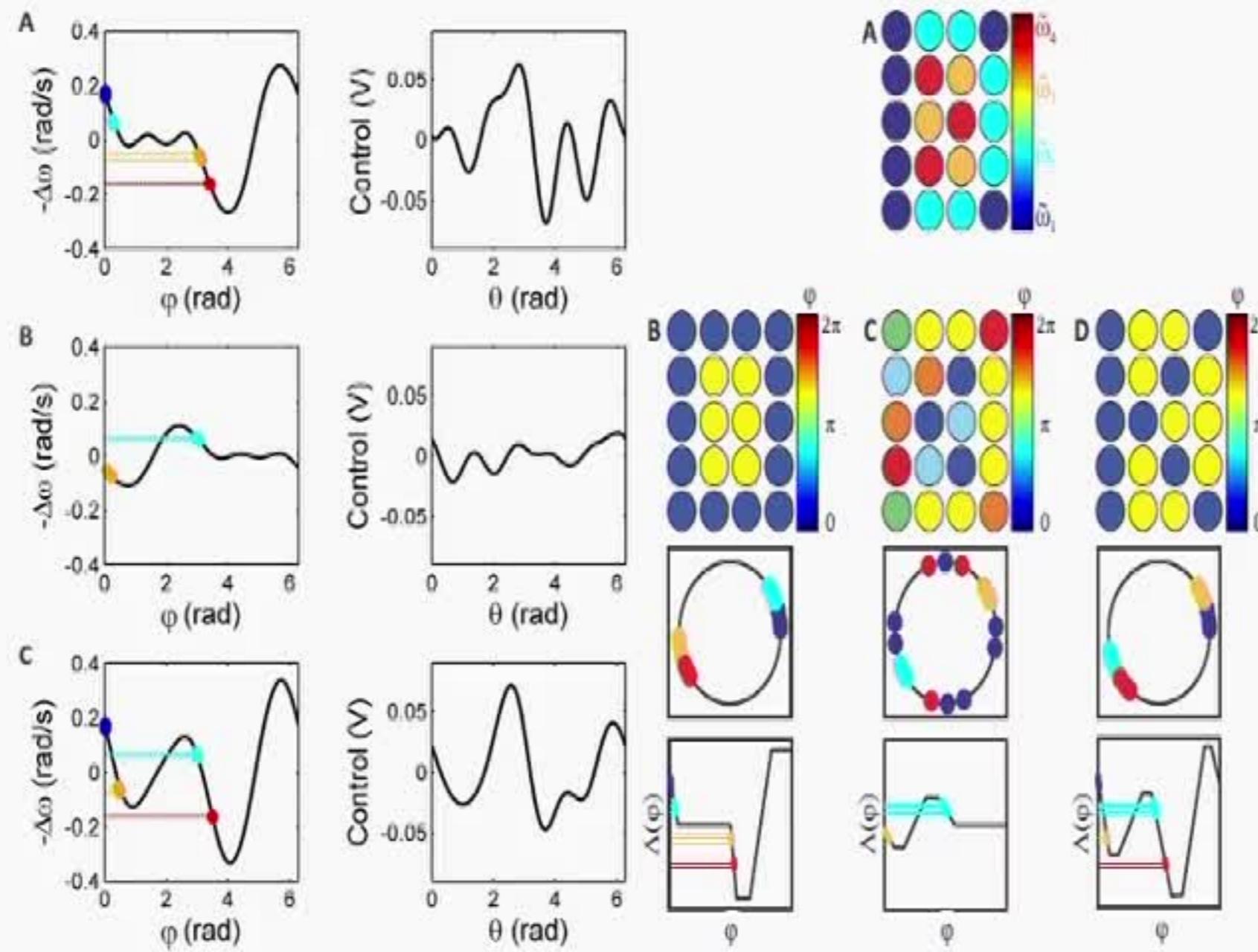
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2 Phase-Selective Entrainment in Oscillator Assemblies

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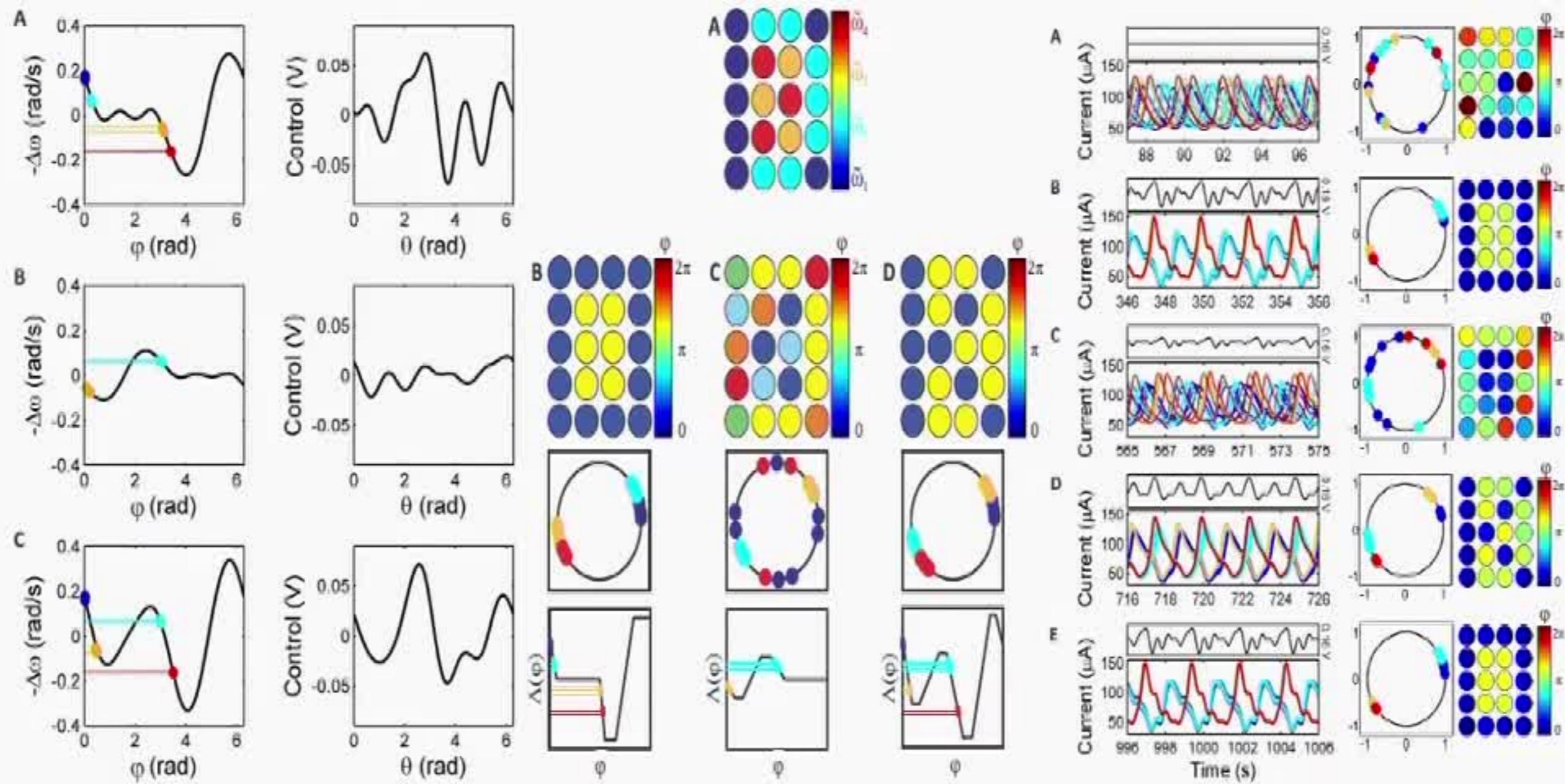
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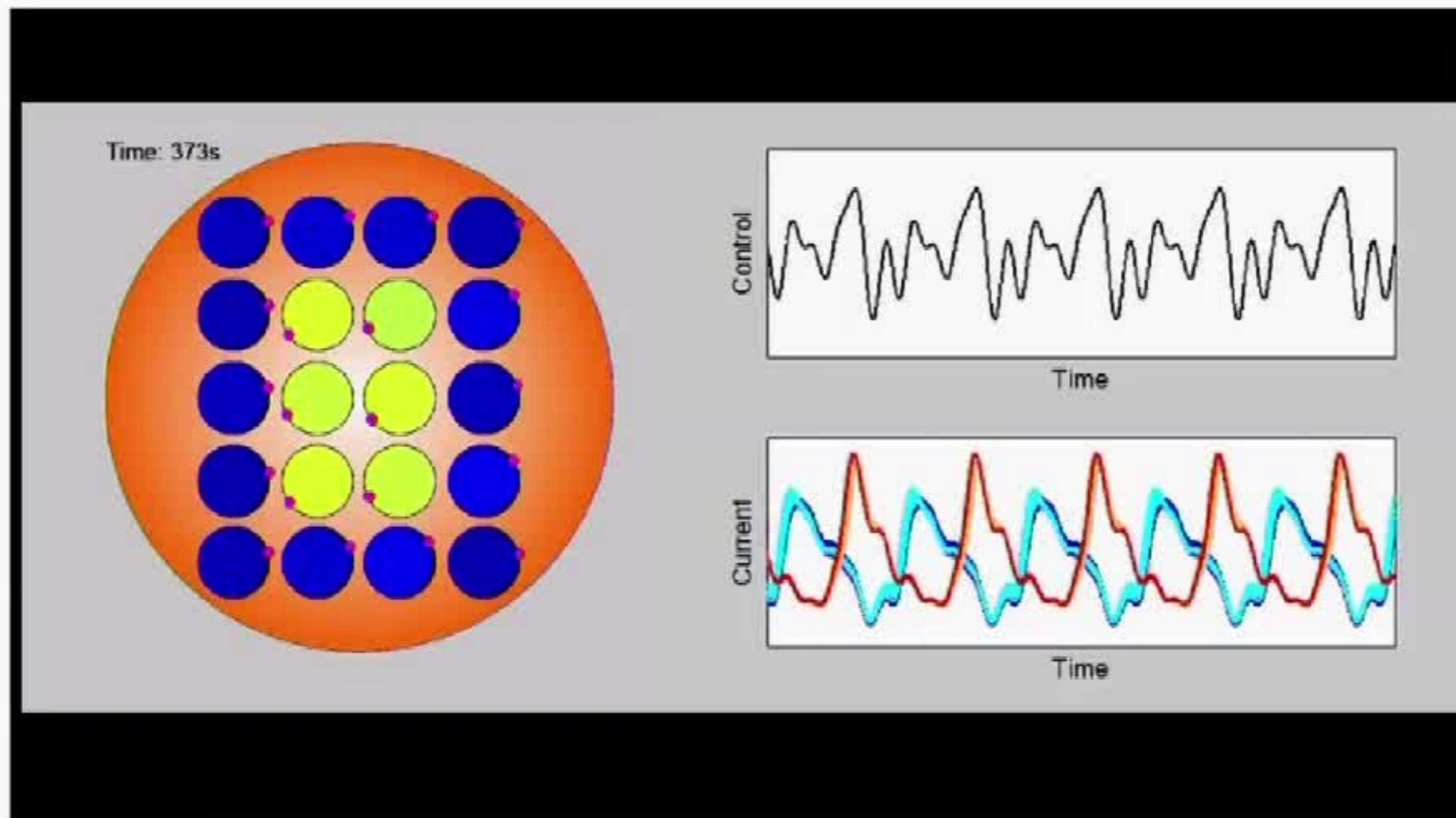


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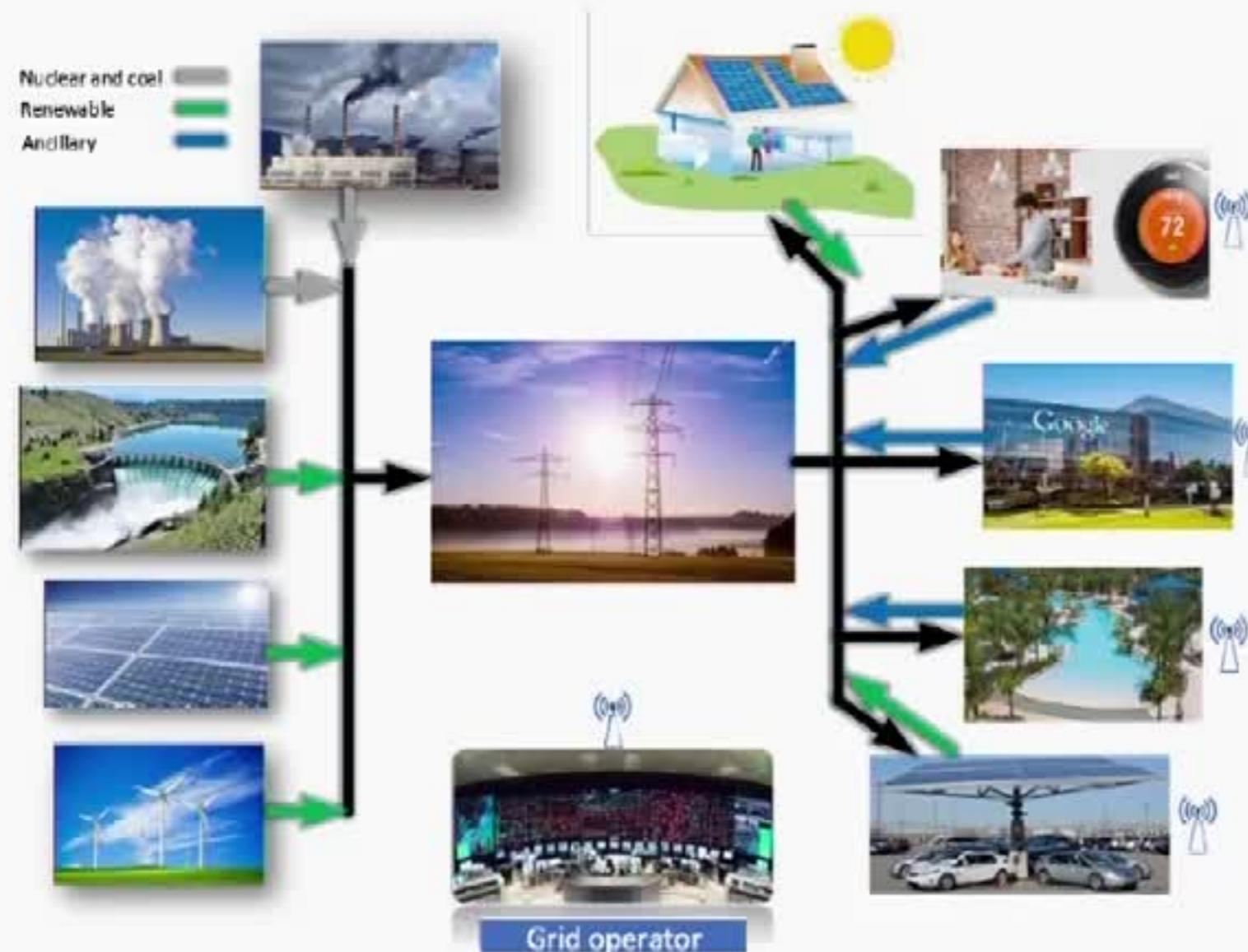
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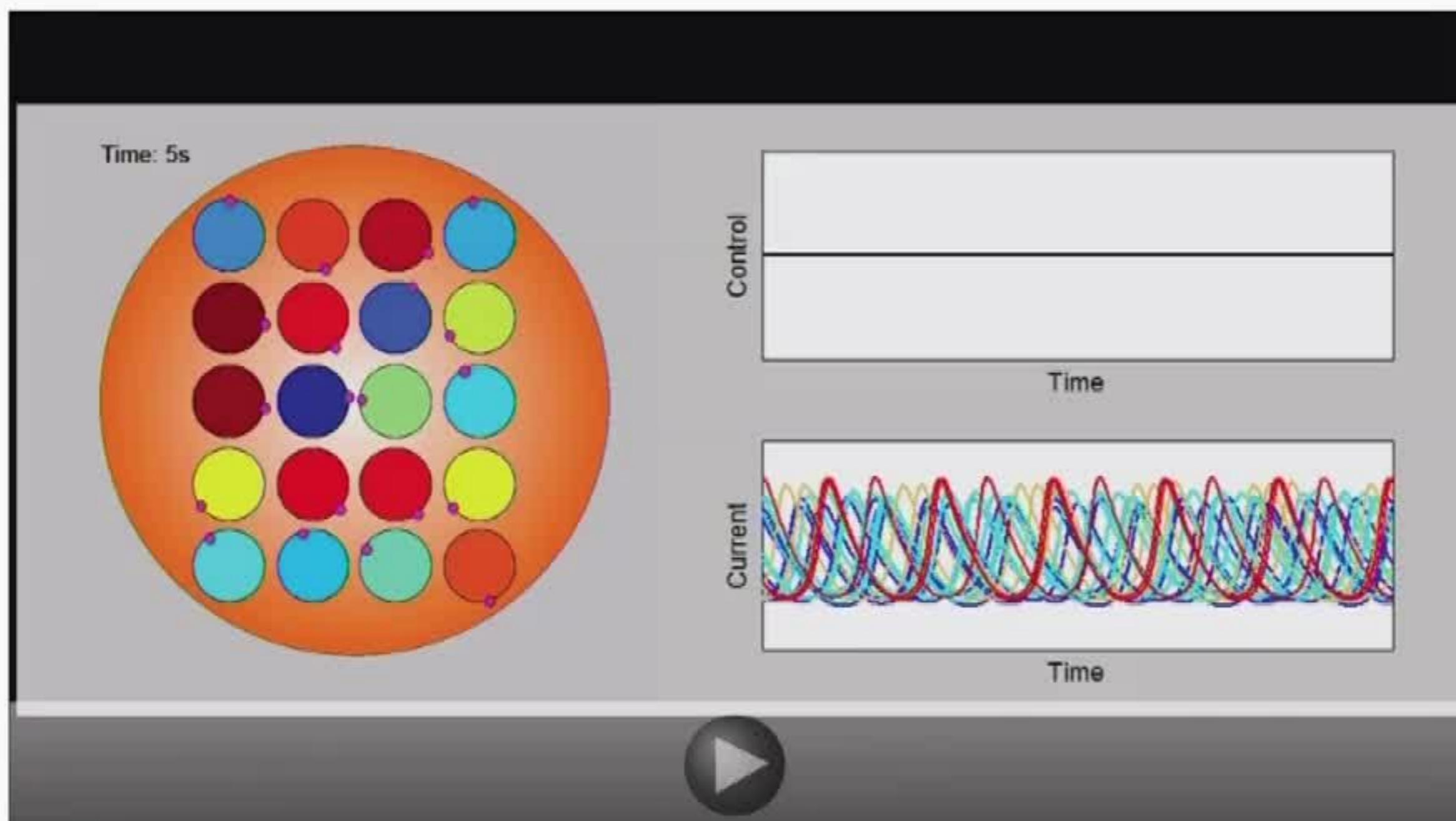
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Demand Response

- Increasing penetration of clean, renewable energy (20% renewables by 2030)



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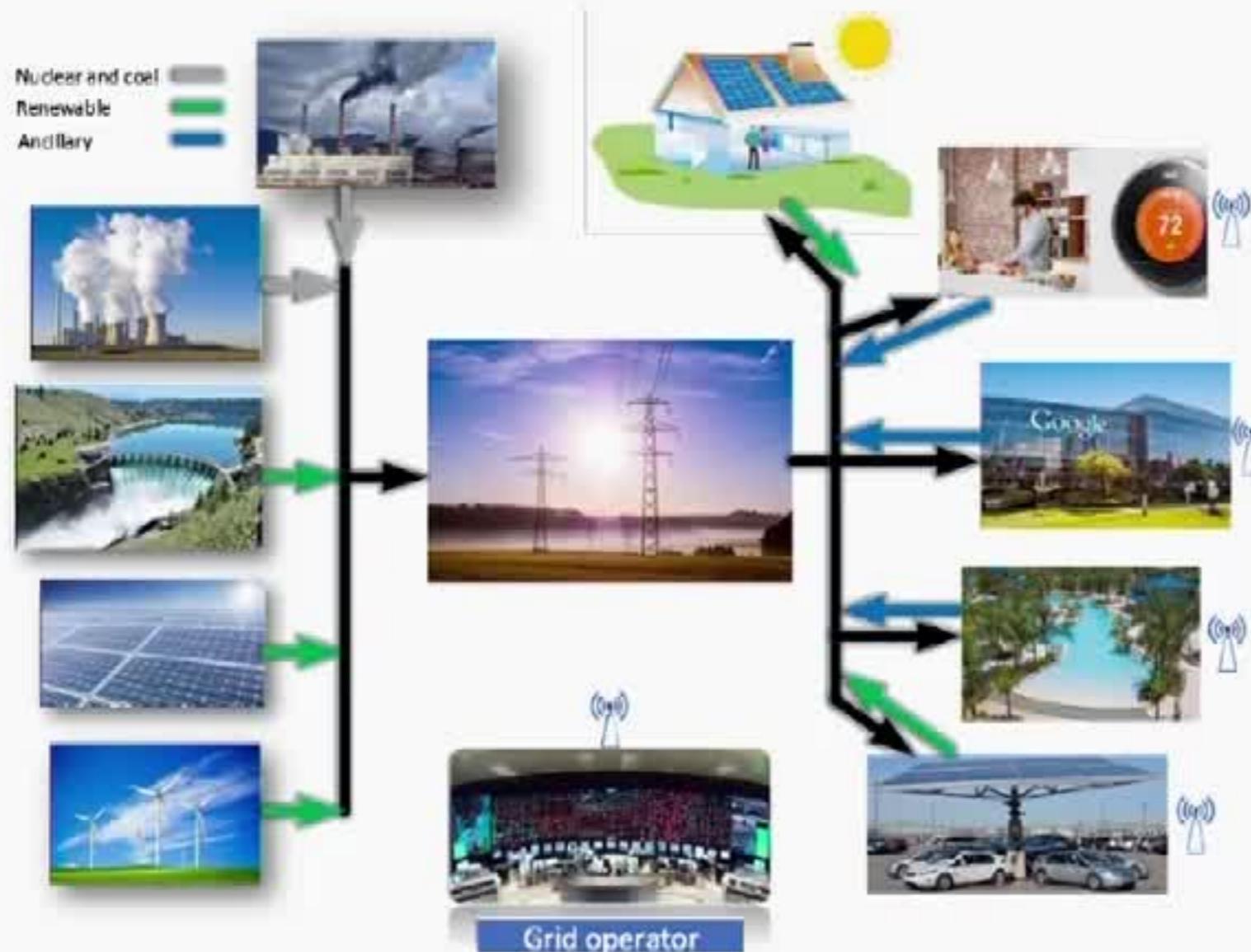
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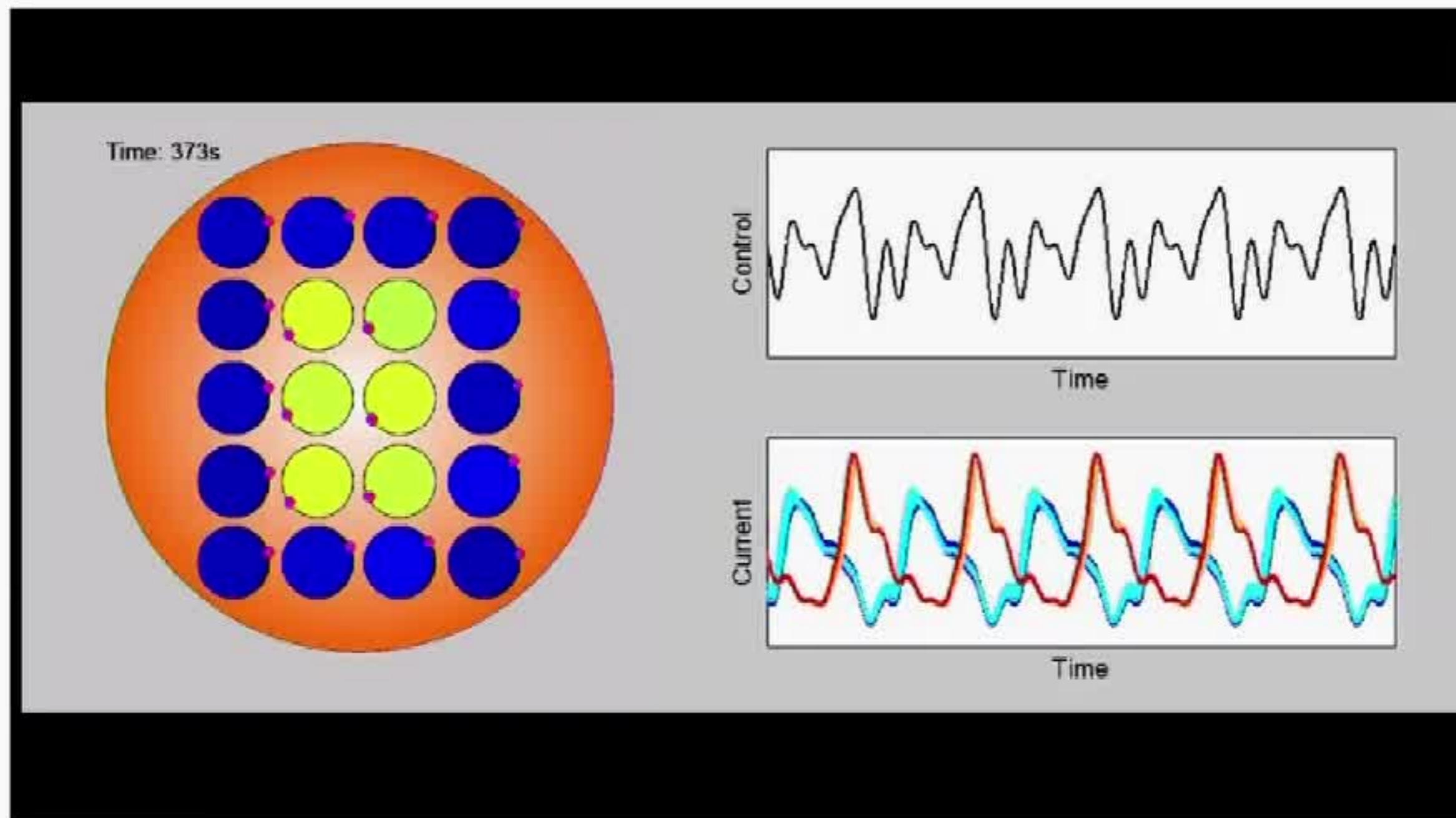
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- Distributed generation, microgrids, new technology - “smart grid”
- Demand Response** provides “ancillary services” to balance loads on the grid

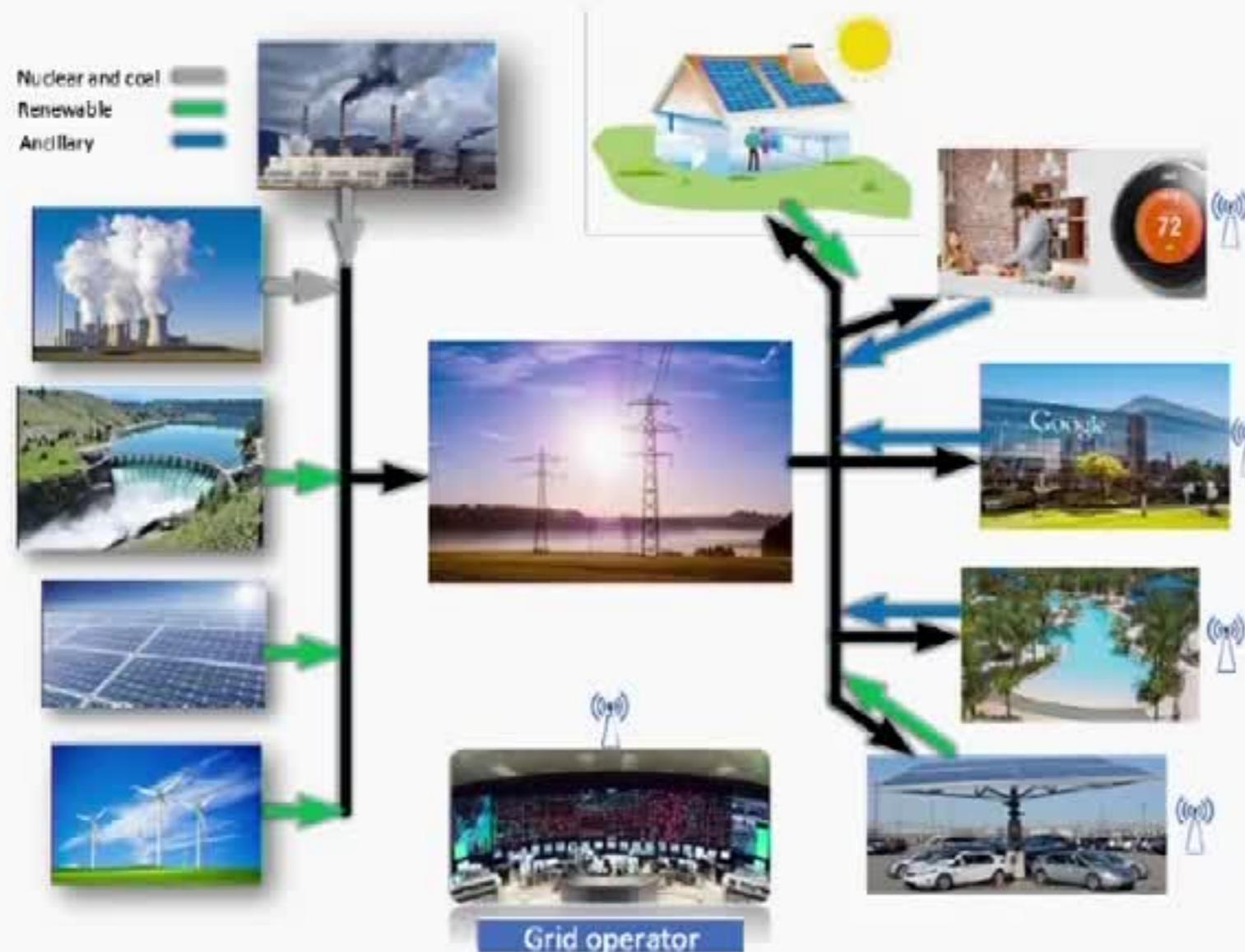
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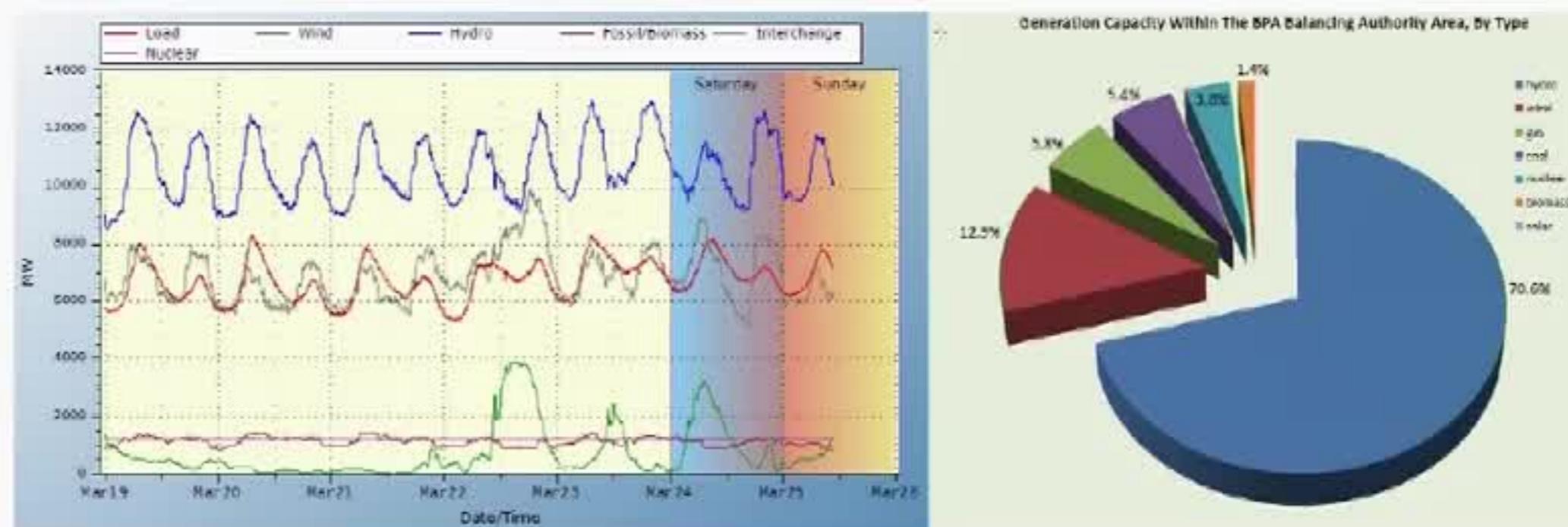
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Demand Response

- Bonneville Power Administration Balancing Authority Load and Total Wind, Hydro, Fossil/Biomass, Nuclear Generation, and Net Interchange



- We want responsive regulation using deferable loads
- Thermostatically Controlled Loads (TCLs), e.g. air conditioners, refrigerators
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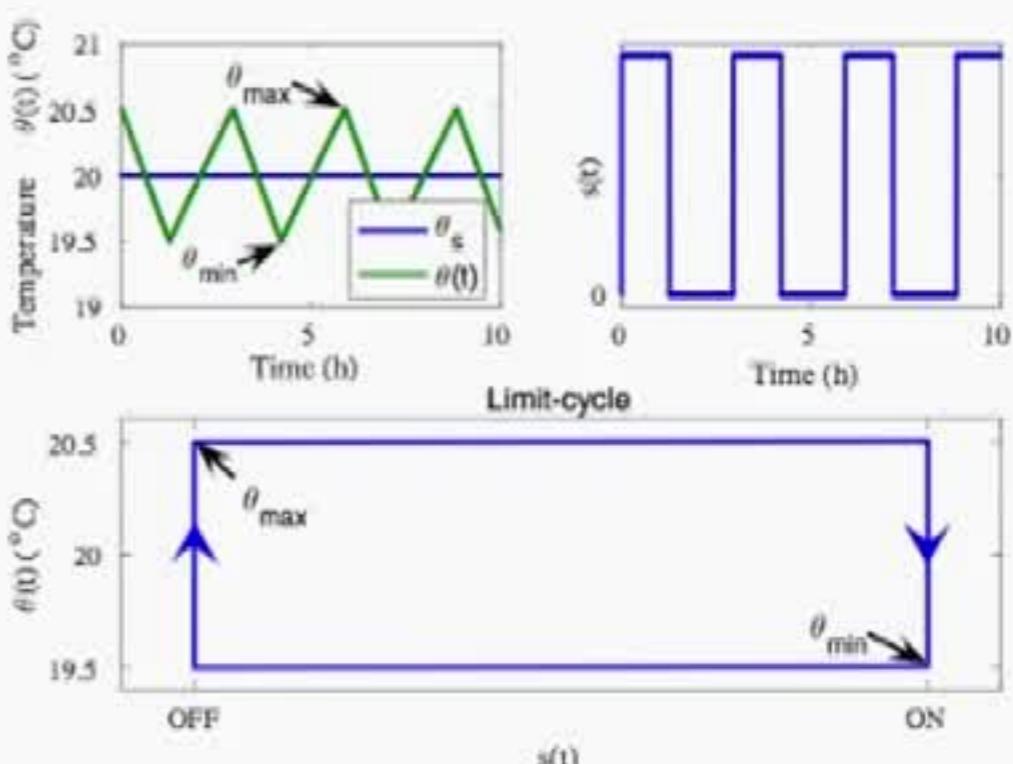
Modeling TCL as an Oscillator

- Consider air conditioner units
- Hybrid-states dynamical model

$$\dot{\theta}(t) = -\frac{1}{RC}[\theta(t) - \theta_a + s(t)PR]$$

$$s(t) = \begin{cases} 0 & \text{if } \theta(t) < \theta_{\min} \\ 1 & \text{if } \theta(t) > \theta_{\max} \\ s(t) & \text{otherwise} \end{cases}$$

Parameter	Meaning	Value
θ_s	temperature setpoint	20°C
θ_a	ambient temperature	32°C
δ	thermostat deadband	1.5°C
R	thermal resistance	2°C/kW
C	thermal capacitance	1.8 kWh/°C
P	energy transfer rate	14 kW
η	coefficient of performance	2.5



- Oscillation period

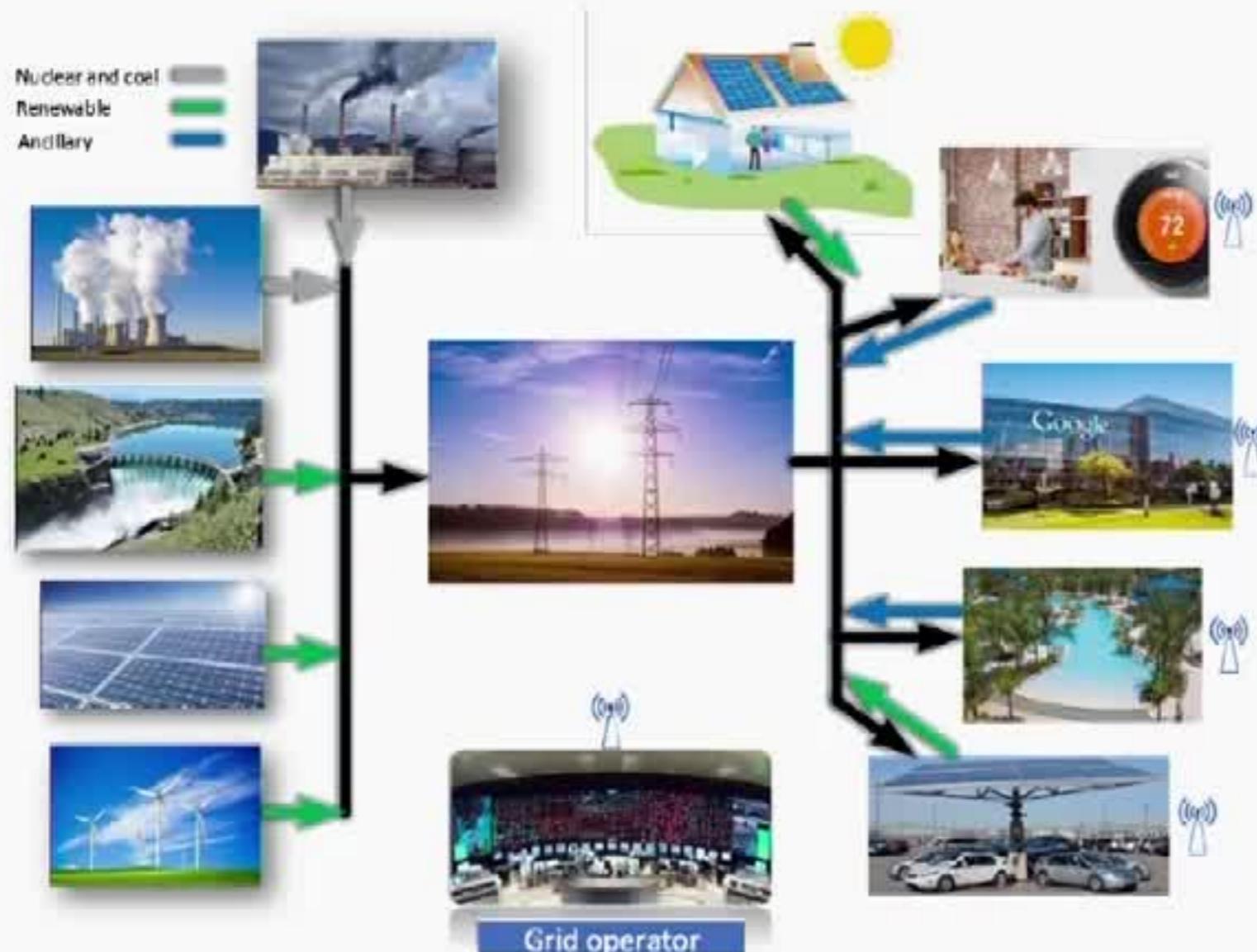
$$T = T_{\text{on}} + T_{\text{off}}$$

$$T_{\text{on}} = RC \ln \left(\frac{\theta_{\max} - \theta_a + PR}{\theta_{\min} - \theta_a + PR} \right)$$

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Demand Response

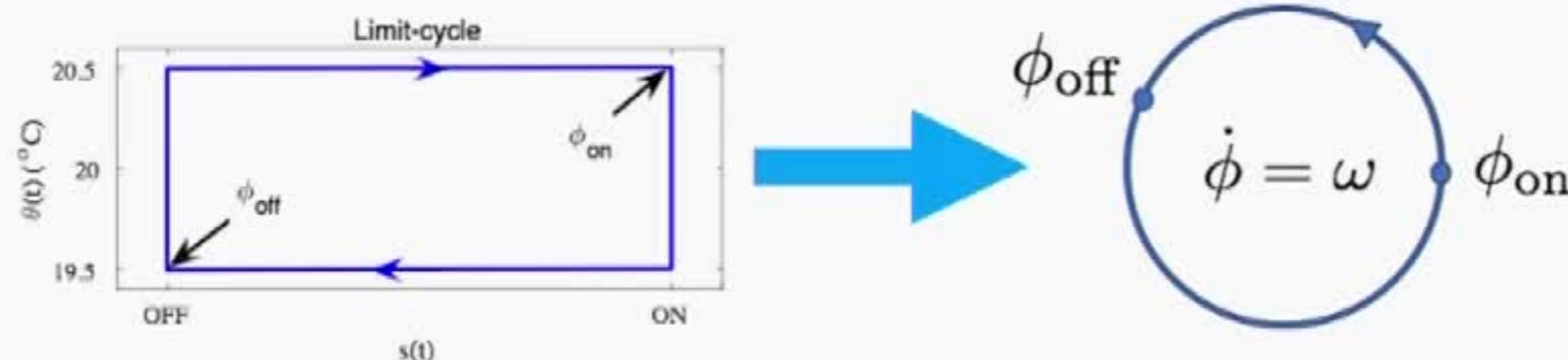
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Modeling TCL as an Oscillator

- Use phase coordinate transformation

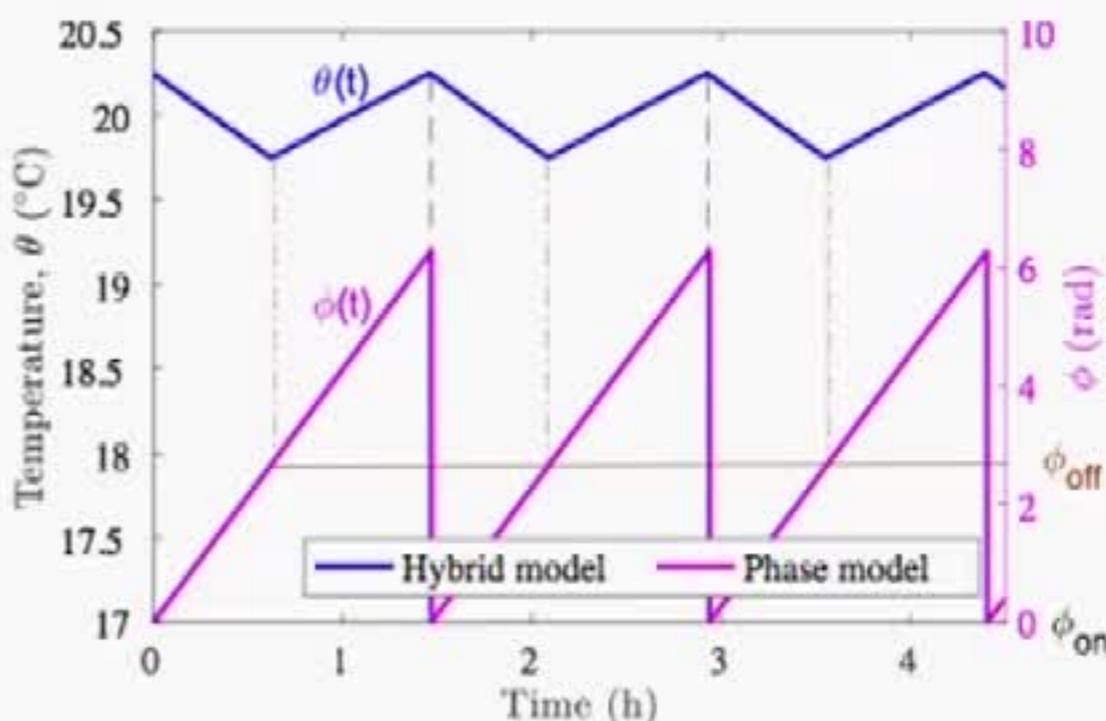


- Controlled phase model

$$\frac{d\phi}{dt} = \omega + Z(\phi)u(t)$$

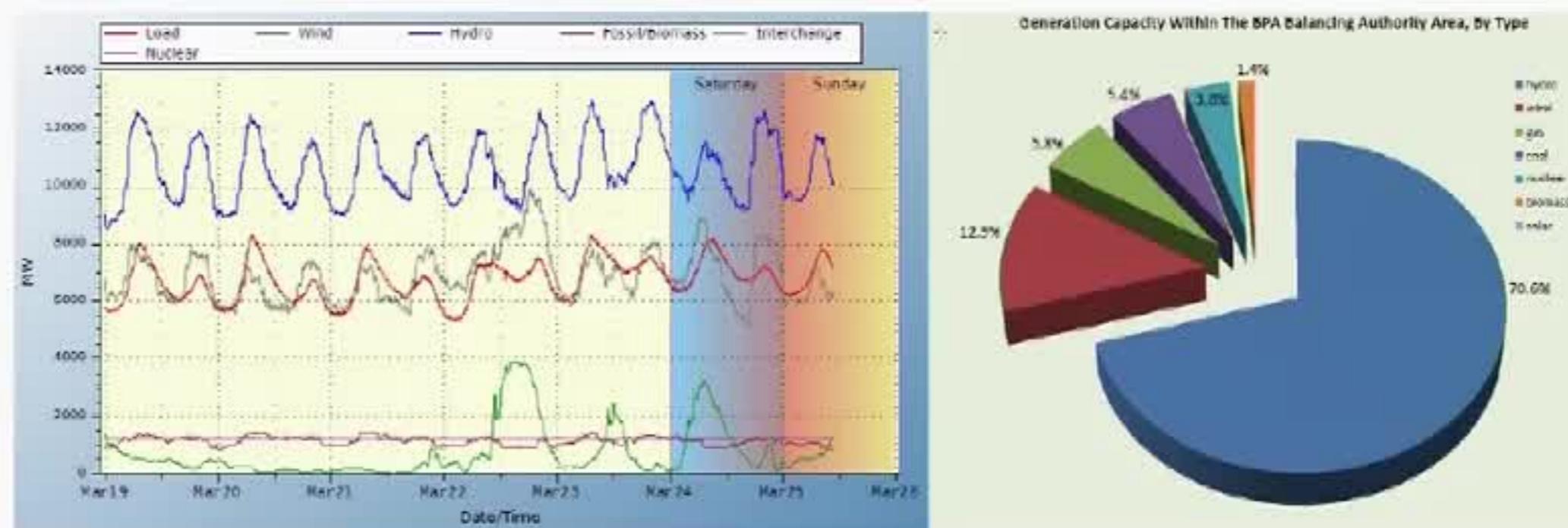
$\omega = \frac{2\pi}{T}$: free running frequency

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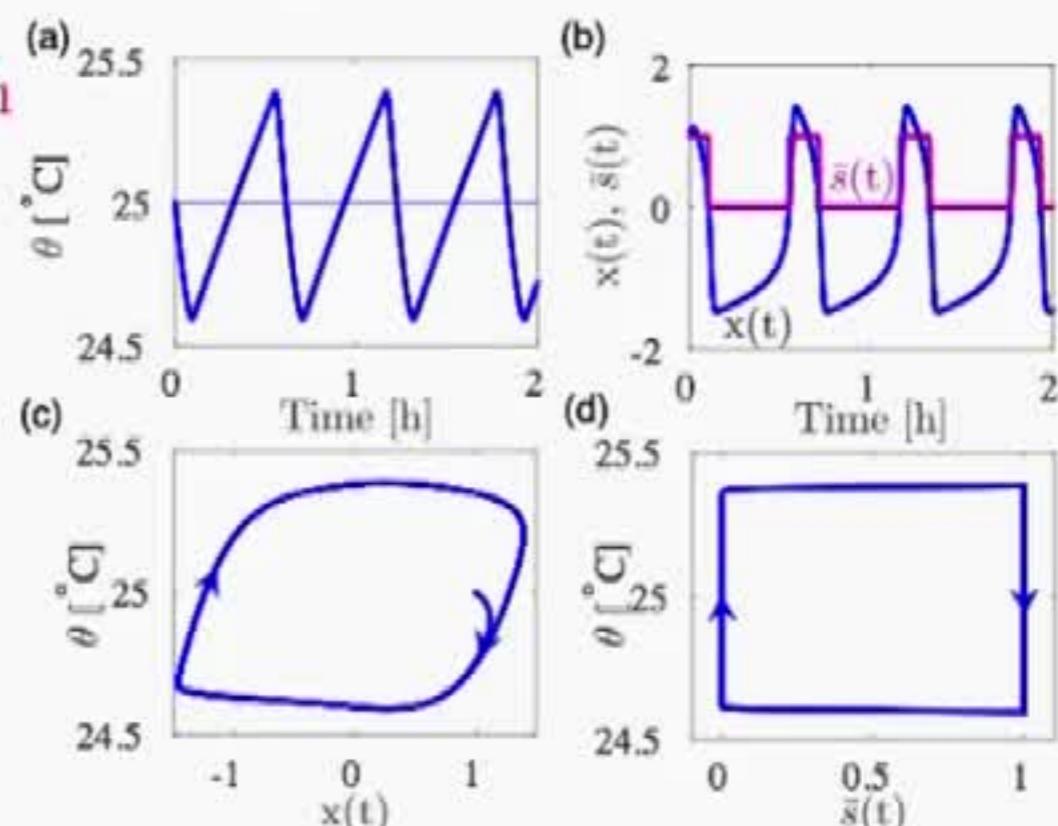
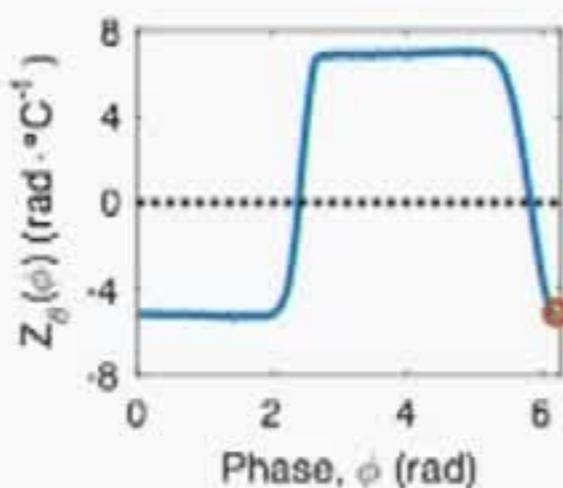
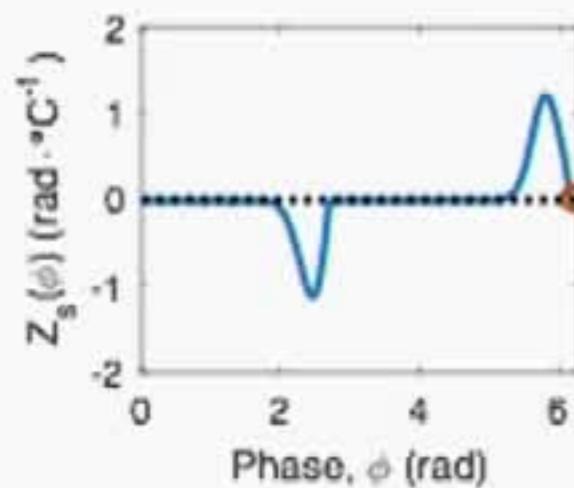
- 2-D TCL continuous model (FitzHugh-Nagumo neuron model)*

$$\dot{x}(t) = \mu \left(\left(\frac{\delta}{2} + \sigma \right) x - \frac{x^3}{3} + \theta - \theta_s \right) + v_1$$

$$\dot{\theta}(t) = -\frac{1}{RC} (\theta - \theta_a + \bar{s}(t) PR) + v_2$$

- Heaviside step function smooth approximation

$$\bar{s}(t) = \frac{1}{2}(1 + \tanh(kx))$$



PRCs

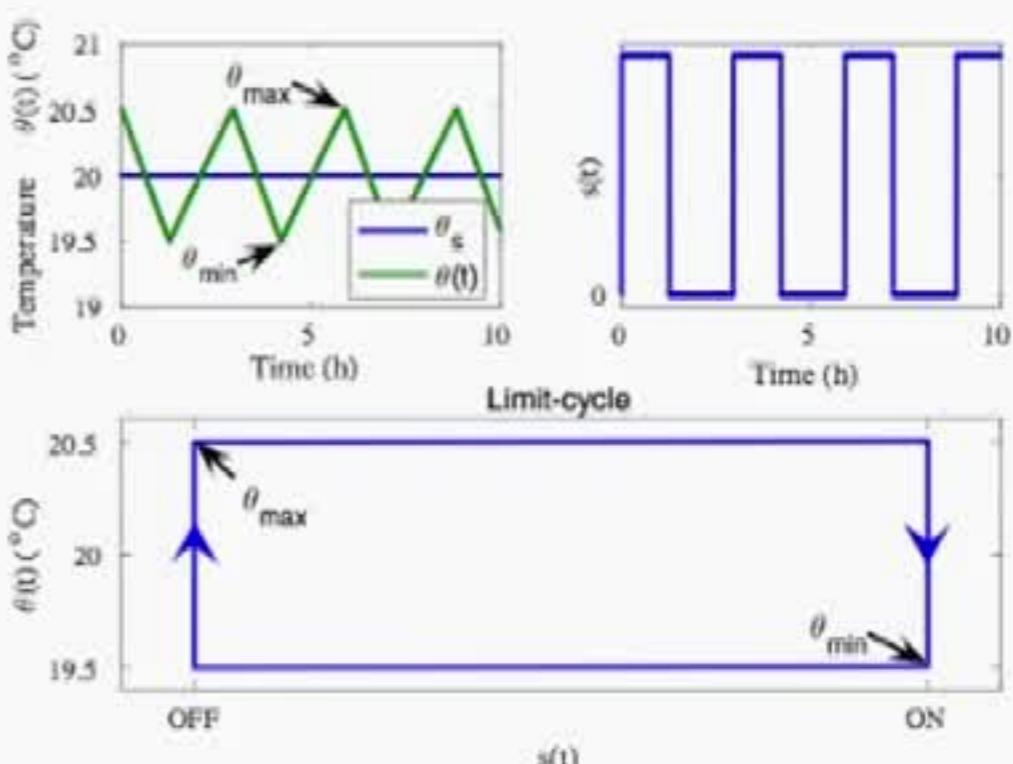
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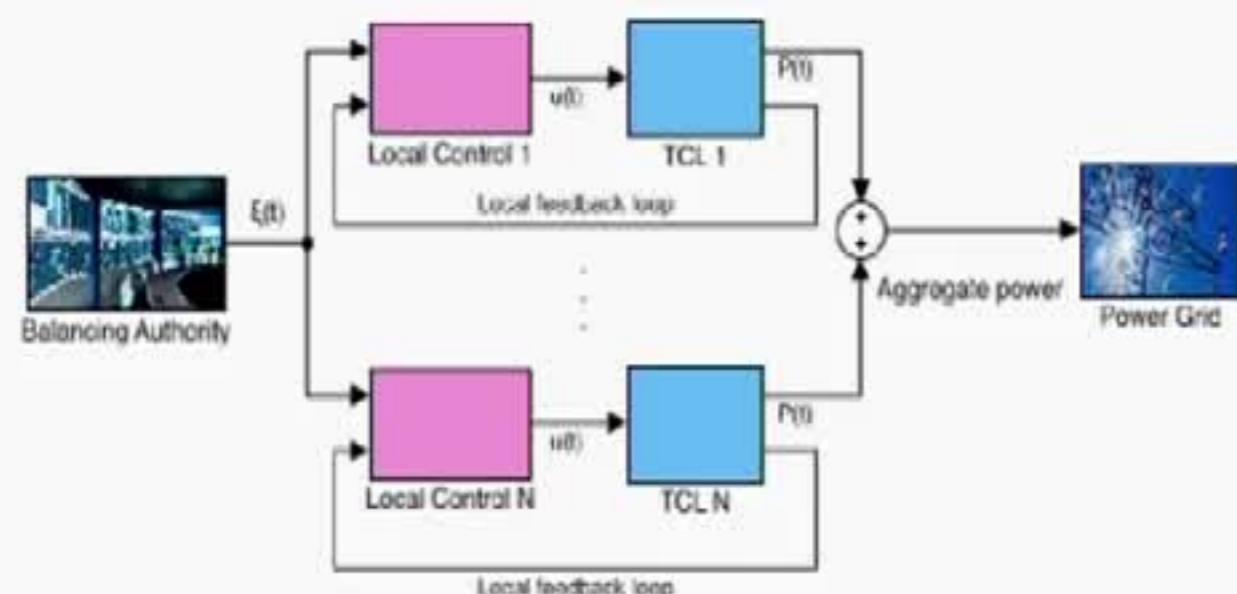
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Control design for TCL ensemble regulation

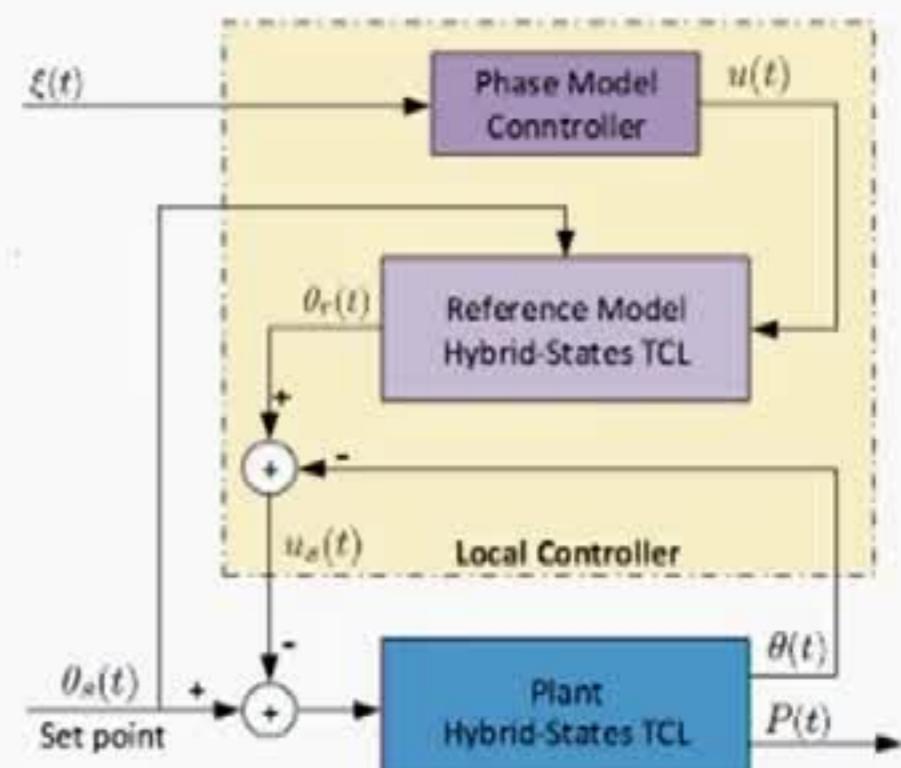
- We want to track the regulation signal for modulating the aggregate power, sent from the balancing authority

- Open-loop control



- Phase model based local controllers

$\xi(t)$: Power regulation signal
 $u(t)$: Control signal to ref. model
 $u_s(t)$: Control signal to the plant
 $\theta_r(t)$: Ref. temperature
 $\theta(t)$: TCL temperature
 $\theta_s(t)$: ste point temperature

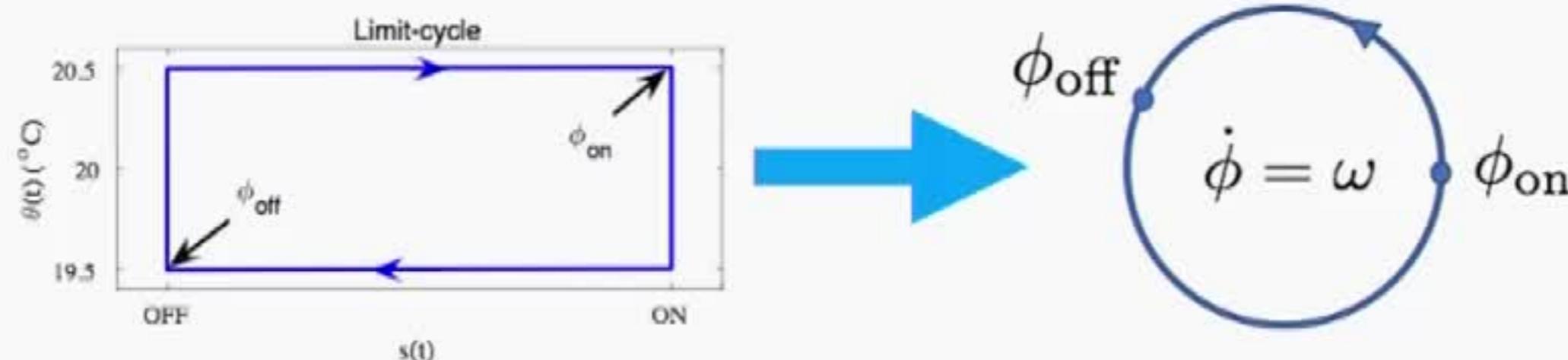


$$\dot{u}(t) = I_1 [I_2 \text{sgn}(\frac{1}{2} - s(t)) Z(\phi(t)) \xi(t) - u(t)]$$

$$u_{k+1} = u_k + I_1 h [I_2 \text{sgn}(\frac{1}{2} - s_k) Z(\phi_k) \xi_k - u_k]$$

Modeling TCL as an Oscillator

- Use phase coordinate transformation

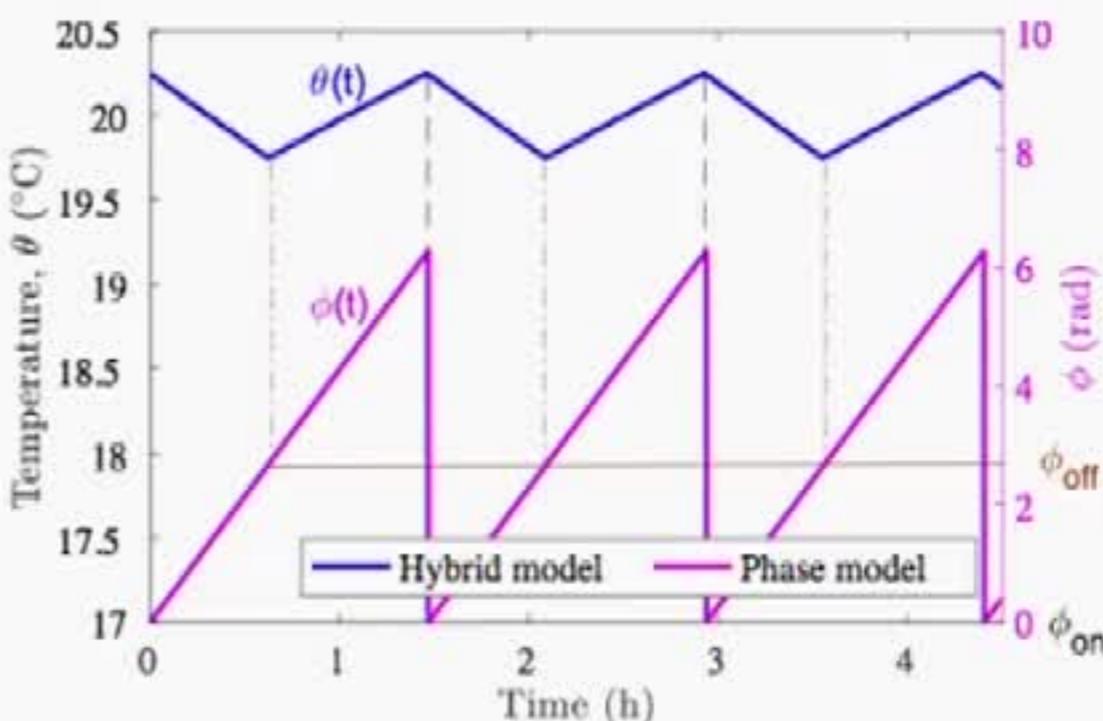


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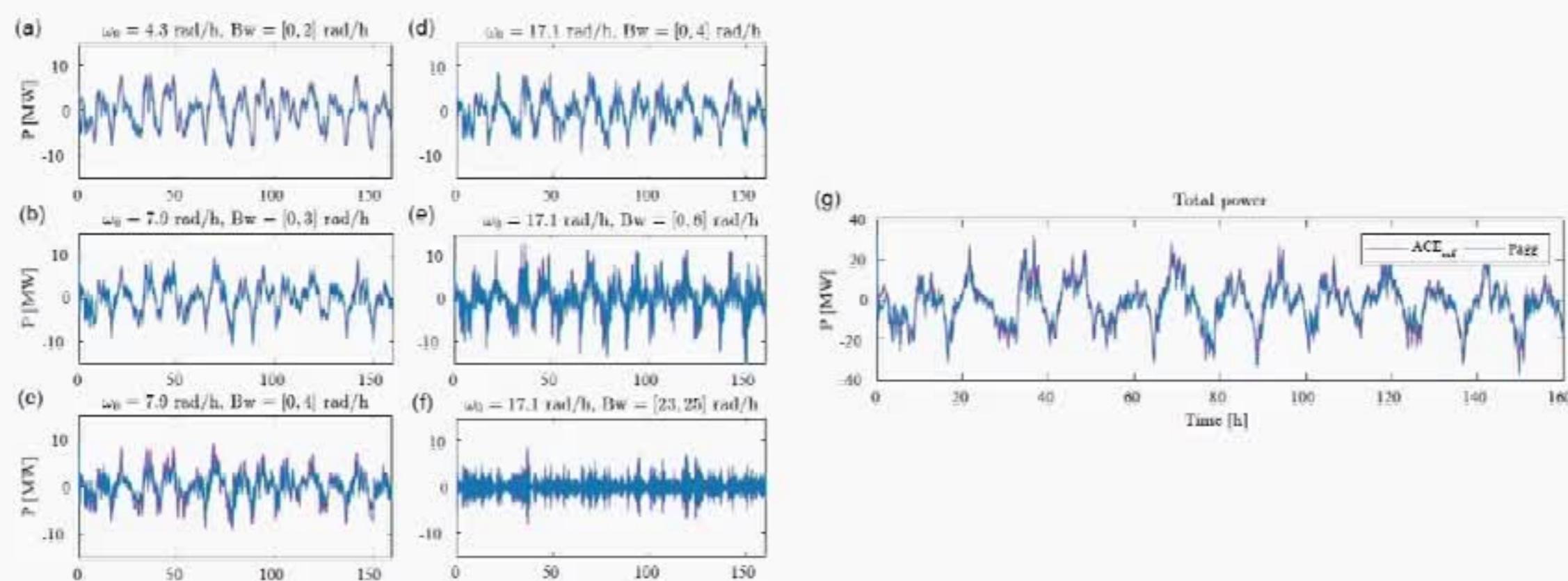
$\omega = \frac{2\pi}{T}$: free running frequency

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Modeling TCL as an Oscillator

- Filter the regulation signal by frequency bands, and use loads with different natural cycling frequencies to track each band



Modeling TCL as an Oscillator

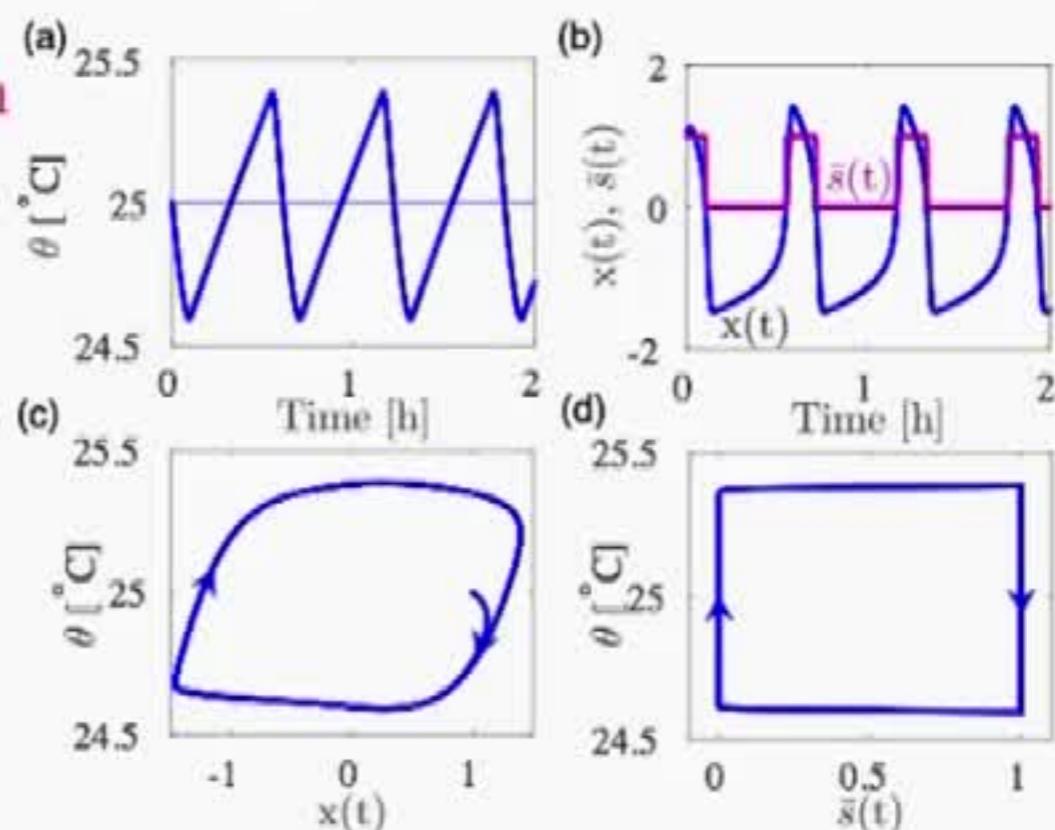
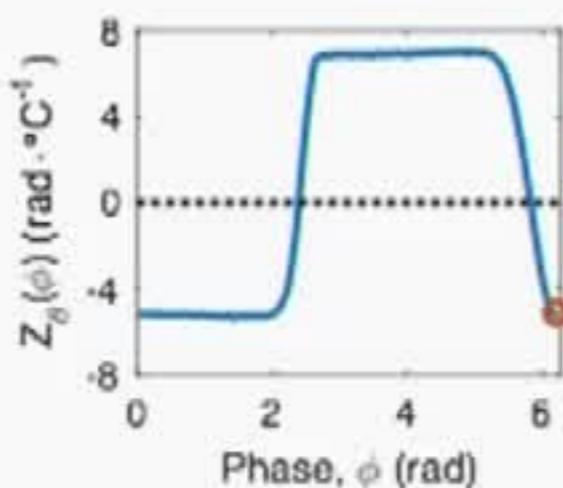
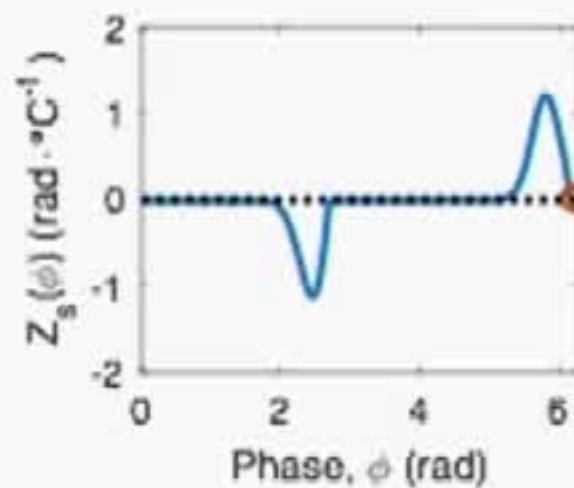
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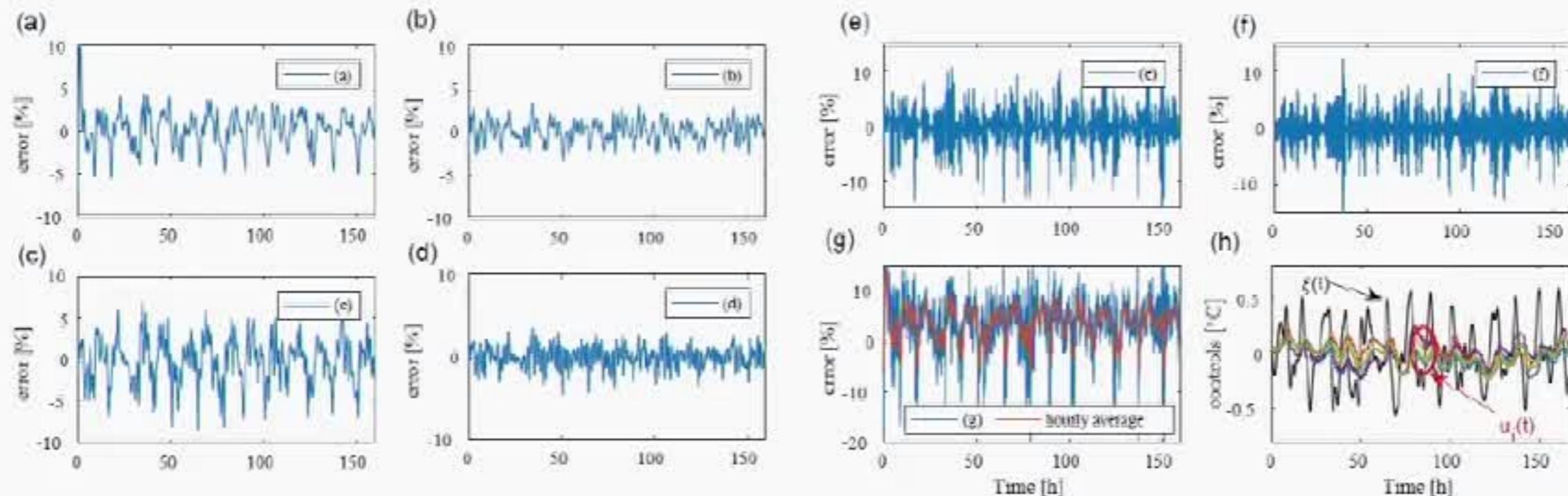
$$\bar{s}(t) = \frac{1}{2}(1 + \tanh(kx))$$



PRCs

Modeling TCL as an Oscillator

- Good tracking performance for each frequency band, and overall < 6% RMSE



$$\text{RMSE \%} = \sqrt{\frac{1}{T} \int_0^T (P_{ref}(t) - P_{agg}(t))^2 dt} \times 100$$

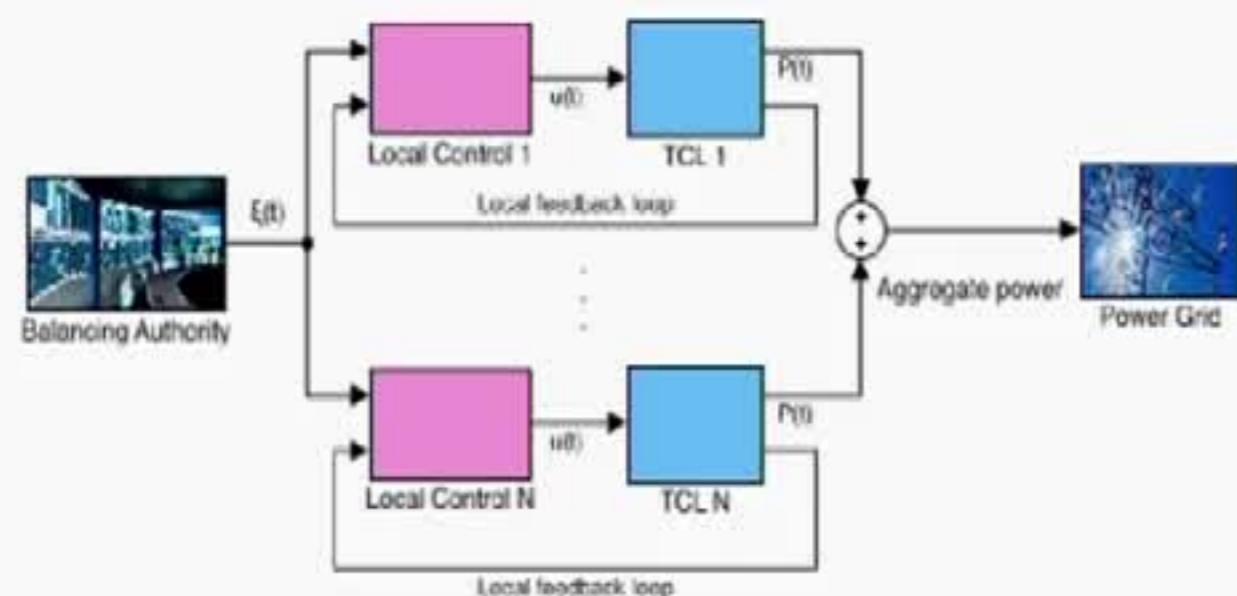
TCL groups	(a)	(b)	(c)	(d)	(e)	(f)	(g)
P_{max} [MW]	10.3	11.2	10.5	9.85	16.5	8.72	39.5
RMSE %	2.31	1.92	3.09	1.38	2.02	3.39	5.89

Bomela, Walter, Anatoly Zlotnik, and Jr-Shin Li. "A phase model approach for thermostatically controlled load demand response." Applied Energy 228(15), 667-680 (2018).

Control design for TCL ensemble regulation

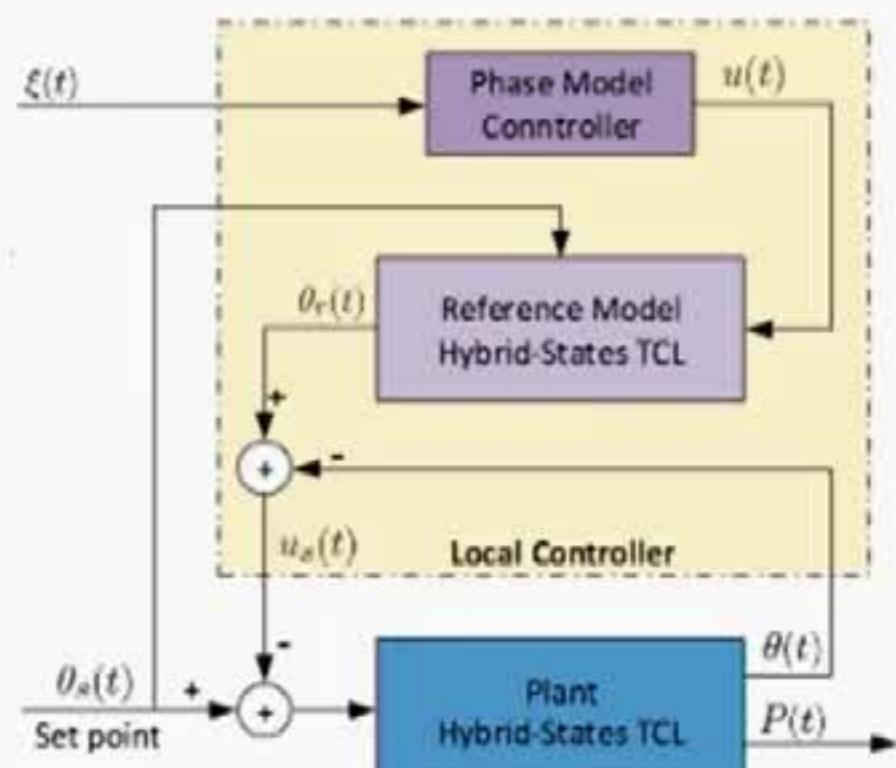
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 $u(t)$: Control signal to ref. model
 $u_s(t)$: Control signal to the plant
 $\theta_r(t)$: Ref. temperature
 $\theta(t)$: TCL temperature
 $\theta_s(t)$: ste point temperature

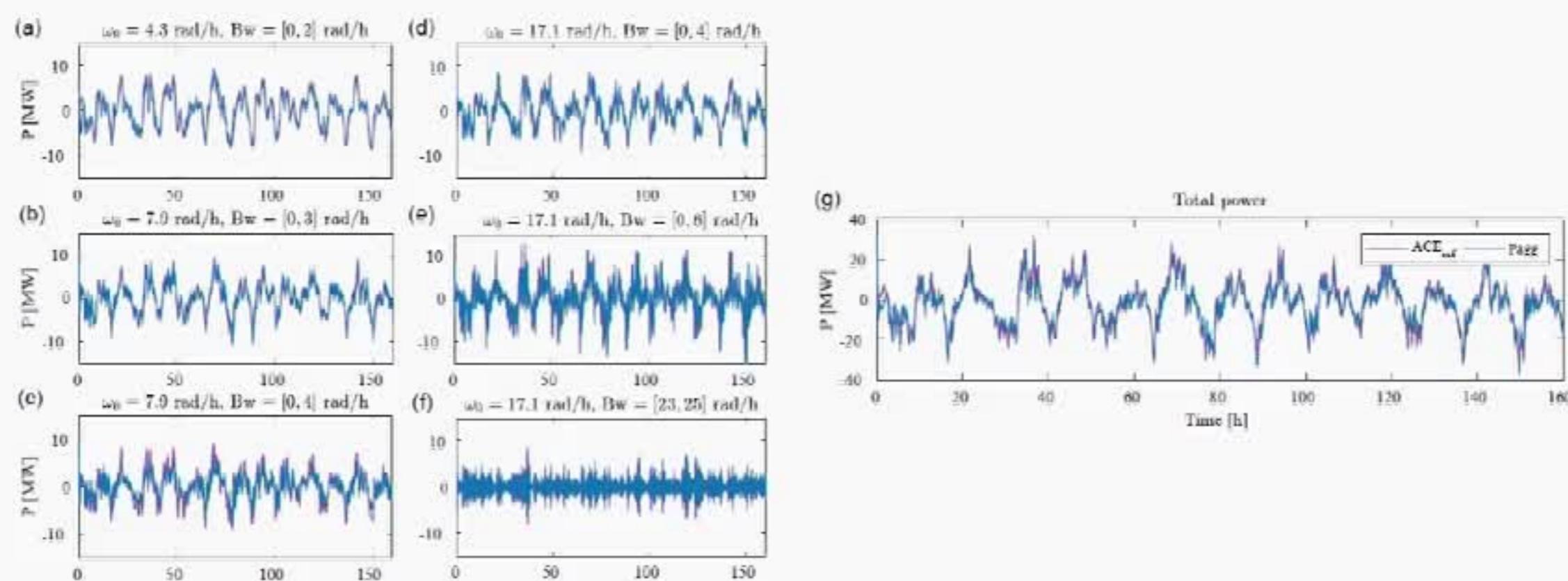


$$\dot{u}(t) = I_1[I_2 \text{sgn}(\frac{1}{2} - s(t)) Z(\phi(t)) \xi(t) - u(t)]$$

$$u_{k+1} = u_k + I_1 h [I_2 \text{sgn}(\frac{1}{2} - s_k) Z(\phi_k) \xi_k - u_k]$$

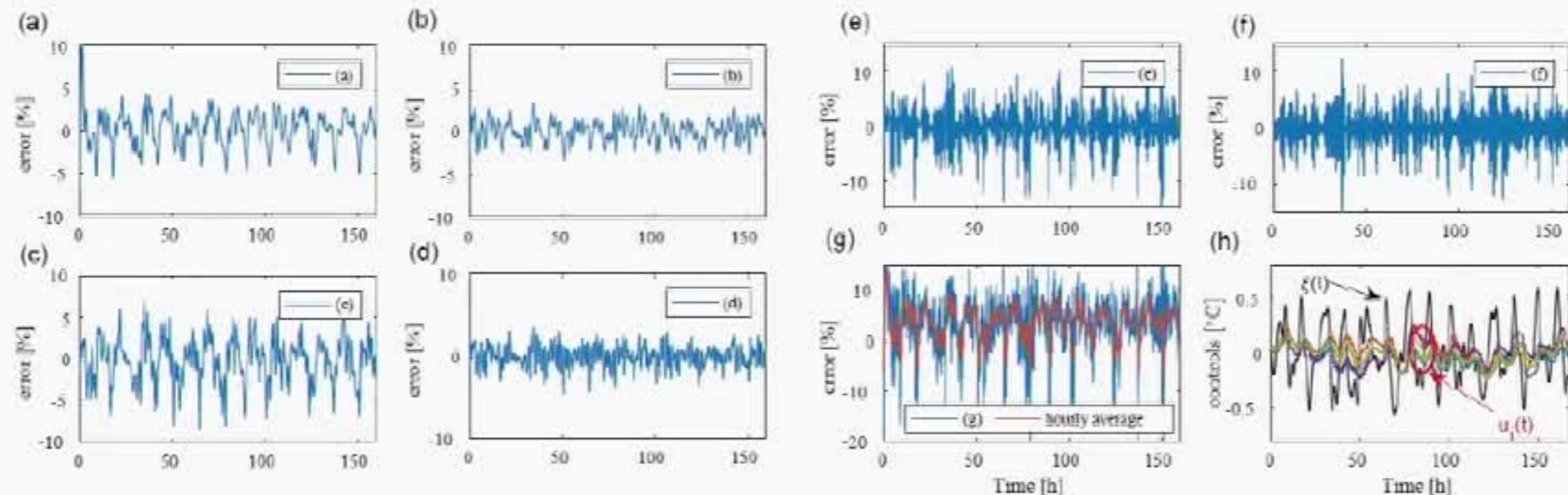
Modeling TCL as an Oscillator

- Filter the regulation signal by frequency bands, and use loads with different natural cycling frequencies to track each band



Modeling TCL as an Oscillator

- Good tracking performance for each frequency band, and overall < 6% RMSE



$$\text{RMSE \%} = \sqrt{\frac{1}{T} \frac{\int_0^T (P_{ref}(t) - P_{agg}(t))^2 dt}{P_{base}^2}} \times 100$$

TCL groups	(a)	(b)	(c)	(d)	(e)	(f)	(g)
P_{max} [MW]	10.3	11.2	10.5	9.85	16.5	8.72	39.5
RMSE %	2.31	1.92	3.09	1.38	2.02	3.39	5.89

Bomela, Walter, Anatoly Zlotnik, and Jr-Shin Li. "A phase model approach for thermostatically controlled load demand response." Applied Energy 228(15), 667-680 (2018).

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